Chirp Analyzer for Estimating Non-stationary Auditory Signals

Y. García-Puente¹, P. Prado-Gutiérrez^{1,2} and E. Martínez-Montes¹ ¹Centro de Neurociencias de Cuba, La Habana, Cuba ²Centro Interdisciplinario de Neurociencia de Valparaíso, Valparaíso, Chile

1 INTRODUCTION

The development of clinical tools for objectively measuring the auditory temporal processing is important for the early diagnosis of speech pathologies related to hearing impairments. In this regards, the use of mathematical models and signal processing techniques is essential to characterize the electrophysiological responses to speech-related auditory stimuli. In this work, we present a Chirp Analyzer (CA), as a new tool for the reliable estimation of non-stationary auditory electrophysiological responses. We study its properties and potential applicability by comparing the estimated responses with those obtained by standard timefrequency methodologies, such as Short Time Fourier Transform and Morlet Wavelet Transform.

2 METHODS

The Envelope Following Response (EFR) is an auditory evoked potential elicited by acoustic stimuli consisting in a carrier tone whose amplitude is modulated by a chirp, i.e. a sinusoidal function with a continuous sweep of amplitude modulation frequencies. The physiological properties of the auditory system suggest that the EFR strongly depend on the modulation signal (chirp). Therefore, the instantaneous estimated amplitude can be considered a measure of the hearing ability to response to each instantaneous modulation frequency (IMF). (Purcell *et al.*, 2004; Prado-Gutierrez *et al.*, 2012).

2.1 Simulated and Real Data

The simulated data consisted of a chirp with IMF linearly varying from 20 to 120 Hz in each half as a reference signal, multiplied by a simulated EFR which imposes instantaneous amplitudes (envelope). Here we simulated the EFR with different shapes and delay (time difference between stimulus onset

and the electrophysiological response), adding in all cases noise with signal-to-noise ratio SNR=2.

As real data, we used electrophysiological recordings of adult rats, obtained in response to amplitude-modulated carrier tones of 4 kHz. Stimuli were delivered at 50 and 70 dB SPL. Chirp was characterized by a linear sweep of IMF from 90 to 200 Hz in each half (15.36 s) of the stimulus. Estimated EFRs were compared with the classical EFR obtained with the Fourier Analyzer (FAM) implemented in the MASTER system (Prado-Gutierrez *et al.*, 2012).

2.2 Short Time Fourier Transform

With the STFT, the non-stationary signal to be analysed is divided into segments that can be considered stationary. This uses a window function $g(t, \tau)$ with fixed temporal width, which implies that the temporal and spectral resolutions are the same in the whole time-frequency plane. The STFT is obtained as the Fourier transform of the product of this window and the signal x(t).

STFT(
$$\tau$$
, f) = $\int x(t)g(t, \tau)e^{-i2\pi ft}dt$ (1)

In this work, we use the Goertzel algorithm to estimate the STFT (discrete version) at predetermined frequencies, using a Hamming window (Boashash, 2003). With this method, the EFR is obtained as the absolute value of the complex coefficients in the time and frequency corresponding to the stimulus' IMF.

2.3 Morlet Wavelet Transform

The continuous wavelet transform (CWT) is defined by:

$$CWT(\tau, f) = \int x(t)W(f, t - \tau)dt$$
 (2)

Where, in our case, the function W(f, t) is the Morlet "mother wavelet":

$$W(f,t) = (\sigma_t \sqrt{\pi})^{-1/2} e^{-t^2/2\sigma_t^2} e^{i2\pi ft}$$
(3)

where the temporal support σ_t is inversely proportional to the spectral support σ_f (Boashash, 2003). The magnitude $z = f/\sigma_f$ is kept constant. Therefore, the spectral resolution is lower and the temporal resolution is higher for increasing values of IMF. This property makes CWT more attractive than STFT for analyzing transient high-frequency phenomena. The EFR is then extracted as the absolute values of the wavelet complex coefficients in times and frequencies corresponding to IMFs.

2.4 Chirp Analyzer

Instead of using a Fourier Basis, the chirp analyzer (CA) proposed here consists in correlating the signal x(t) with a non-stationary reference function $\varphi(t)$ that represents the theoretical response:

$$CA(\tau) = \int x(t)g(t,\tau)\phi(t)dt \qquad (4)$$

This procedure is carried out in overlapping rectangular windows $g(t, \tau)$, for achieving a higher temporal precision with the same spectral resolution. As this is done directly in the time domain, this method is faster than the other methods that need to estimate coefficients for all frequencies in each time point. However, this also makes this method sensitive to the phase difference between the signal and the reference function, since correlation vanishes when the signals are in counterphase. The estimated EFR is extracted from values of CA in the time points corresponding to each IMF.

3 RESULTS

Figure 1 shows the response estimated by the three methods (blue lines) for the simulated signals (red lines) with different shapes, varying smoothness and modulation depth. In all cases, the signals were simulated with a peak signal-to-noise ratio SNR = 2 (noise variance is the half of signal's maximum). The first shape (left column) corresponds to a sinusoidal squared response, with 100% modulation depth; the second (central column) is a sinusoid with 50% modulation depth and the third (right column) is a rectangular pulse with 100% modulation depth.

In real conditions, the auditory system responds with a particular time delay with respect to the stimulus onset, known as the latency of the response. Moreover, the electronic equipment used for recording the signals can introduce delays of up to 100 ms in some cases. Given that the methods estimate the amplitude of the response at a particular time, it is important to study how this delay affects the estimated EFR with respect to the one obtained using the exact latency of the response (hereinafter called non-delayed response).



Figure 1: EFR estimation (blue line + dots) from simulated signals with different shapes (red lines).

Figures 2 and 3 show the EFR estimated by the three methods from data simulated with different values of the response's delay.



Figure 2: EFR estimated (red lines) with the CWT and STFT from simulated signals with different values of the auditory response's delay, in comparison with the EFR estimated for the non-delayed response (black lines).



Figure 3: EFR estimated (red lines) with the CA from simulated signals with different values of the auditory response's delay, in comparison with the EFR estimated for the non-delayed response (black lines).

In the analysis of electrophysiological recordings in adult rats, the EFRs were obtained with the three methods, as shown in Figure 4. For comparison purposes, we plotted them normalized, together with the EFR estimated with the Fourier Analyzer (FA) implemented in the stimulation and recording system described in (Purcell *et al.*, 2004; Prado-Gutierrez *et al.*, 2012). The bottom right panel of figure 4 shows the estimated EFR without normalization, which allows comparing the responses' amplitude obtained with each method.



Figure 4: EFR estimated from the real electrophysiological recordings in adult rats. Top row and bottom left: responses were normalized such that all maxima coincide with that of the classical Fourier Analyzer (FA). Bottom right: non-normalized responses estimated with the three methods.

4 DISCUSSION

Although the three methods were able to recover the shape of the simulated responses, some interesting differences were evident (Figure 1). CWT and CA estimated the amplitude with higher accuracy, but the former is more sensitive to noise and therefore, overestimates the amplitude of small or null responses. This is explained by the higher temporal (and lower spectral) resolution of the CWT in the frequency band studied. The STFT did not estimate the amplitude correctly (underestimating it), due to the violation of the main assumption of this method, i.e. the stationarity of the signal (Boashash, 2003). Also, the rectangular pulse showed that the CA reflected the abrupt changes in the response with slightly lower temporal resolution than the CWT.

Figures 2 and 3 showed that the EFR estimated with the CWT was the less affected by changes in the response's delay. Again, this is explained by its lower spectral resolution for high frequencies, which leads to similar amplitudes of the response in a wide time-frequency range. Contrarily, the higher spectral resolution of the STFT led to great changes in the amplitude estimated in nearby time-frequency points. The CA is the most affected when the response is estimated by selecting the wrong timefrequency points, since this method rely in correlating the signal with a reference function. The delay corresponds to a phase difference between both signals, which makes the correlation to drop drastically (Figure 3).

Results of the analysis of real data showed that the three methods may be considered as useful tools for the estimation of non-stationary auditory evoked responses. The EFR estimated showed similar shapes than the one obtained with the FA, which is one of the most popular methods to study this type of auditory responses (Purcell et al., 2004; Prado-Gutierrez et al., 2012). The CWT presented higher variability and more local extremes, due to its sensitivity to noise. However, this effect can be ameliorated by smoothing the response in a postprocessing (e.g. we used a 7-point sliding window smoother). Regarding the non-normalized responses, the STFT showed the smallest amplitudes. Also, amplitude of the EFR estimated with the CWT was higher than those of the CA for all frequencies, which suggests the existence of a response delay.

In summary, among the three methods studied here, the CA is the fastest (around 3s against more than 30s each of the other two approaches) and most reliable method to estimate the amplitude of the EFR. However, this method is strongly affected when the latency of the response (together with electronic delays) is high. As this value is usually unknown, this method should be used carefully and new ways of estimating the response's delay have to be the goal of future developments. All these results suggest that the CA is a promising tool to estimate the EFR, although optimal estimation could be achieved with a methodology that combines the good properties of the three techniques.

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