

# Simplified Computation of $l_2$ -Sensitivity for 1-D and a Class of 2-D State-Space Digital Filters Considering 0 and $\pm 1$ Elements

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**Abstract:** A simplified method of computing an improved  $l_2$ -sensitivity measure is developed for state-space digital filters by reducing the number of the Lyapunov equations, and it is expanded into a class of two-dimensional (2-D) state-space digital filters. First, a conventional improved  $l_2$ -sensitivity for state-space digital filters is reviewed and simplified to two novel forms so that the number of the Lyapunov equations is reduced. Next, the resulting method is expanded into a class of 2-D state-space digital filters. Finally, two numerical examples are presented to evaluate more precise (improved)  $l_2$ -sensitivity measures for 1-D and a class of 2-D state-space digital filters by employing the proposed methods.

## 1 INTRODUCTION

In the case when a state-space model is realized from a given transfer function and implemented with a finite binary representation, the truncation or rounding of the coefficients is required to meet the finite-word-length (FWL) constraints. As a result, unacceptable degradation of the characteristics of a recursive digital filter may be caused, and a stable recursive digital filter may be changed to an unstable one. This motivates the study of coefficient sensitivity analysis and its minimization problem. Several methods have been explored to evaluate the coefficient sensitivity of a state-space digital filter and to minimize the coefficient sensitivity (Thiele, 1984; Thiele, 1986; Iwatsuki et al., 1989; Li and Gevers, 1990; Li et al., 1992; Yan and Moore, 1992; Li and Gevers, 1992; Gevers and Li, 1993; Xiao, 1997; Hinamoto et al., 2005; Yamaki et al., 2006). The analysis and minimization problems of  $l_2$ -sensitivity have also been considered for two-dimensional (2-D) state-space digital filters (Kawamata et al., 1987; Hinamoto et al., 1990; Li, 1997; Li, 1998; Hinamoto et al., 2002; Hinamoto et al., 2006; Yamaki et al., 2007). Some of them evaluate the coefficient sensitivity by using a mixture of  $l_1/l_2$  norms (Thiele, 1984; Thiele, 1986; Iwatsuki et al., 1989; Li and Gevers, 1990; Li et al., 1992; Kawamata et al., 1987; Hinamoto et al., 1990), while the others rely on the use of a pure  $l_2$  norm (Yan and Moore, 1992; Li and Gevers, 1992; Gevers and Li, 1993; Xiao, 1997; Hinamoto et al., 2005; Li, 1997; Li, 1998; Hinamoto et al., 2002; Hinamoto et al., 2006; Yamaki et al., 2006; Yamaki et al., 2007). It is noted that the  $l_2$ -sensitivity measure is more natural and reasonable than the  $l_1/l_2$  mixed sensitivity measure. In (Xiao, 1997), an improved  $l_2$ -sensitivity measure has been presented to evaluate  $l_2$ -sensitivity more precisely when the state-space model contains 0 and  $\pm 1$  coefficients. In (Hinamoto and Doi, 2012), simple  $l_2$ -sensitivity measures have been explored for evaluating the  $l_2$ -sensitivity of canonical forms in 1-D and 2-D separable-denominator state-space digital filters.

In this paper, a simplified method of computing an improved  $l_2$ -sensitivity measure for state-space digital filters is developed by reducing the number of the Lyapunov equations. The resulting method is expanded into a class of two-dimensional (2-D) state-space digital filters reported in (Hinamoto, 2001). This class of 2-D state-space digital filters can be viewed as a dual system of the Fornasini-Marchesini second local state-space model (Fornasini and Marchesini, 1978). First, a conventional improved  $l_2$ -sensitivity for state-space digital filters in

(Xiao, 1997) is reviewed and simplified to two novel forms so that the number of the Lyapunov equations is reduced. The novel contribution exists in two alternative formulations for the Lyapunov equations where two independent variables are replaced by a single independent variable. This enables one to reduce the amount of computations considerably. Furthermore, the possible coefficient values equal to 0 and  $\pm 1$  are considered as special cases. Then the simplified method of computing the improved  $l_2$ -sensitivity measure for 1-D filter is expanded into a class of 2-D state-space digital filters. The analysis is carried out more precisely than that in (Hinamoto et al., 2006) by taking into account 0 and  $\pm 1$  elements in the 2-D state-space digital filter. Finally, two numerical examples are presented to demonstrate the validity and effectiveness of simplified methods for computing more precise (improved)  $l_2$ -sensitivity measures in 1-D and a class of 2-D state-space digital filters.

## 2 REVIEW OF IMPROVED $l_2$ -SENSITIVITY FOR STATE-SPACE DIGITAL FILTERS

Consider a stable, controllable and observable state-space digital filter  $(\mathbf{A}, \mathbf{b}, \mathbf{c})_n$  described by

$$\begin{aligned} \mathbf{x}(k+1) &= \mathbf{A}\mathbf{x}(k) + \mathbf{b}u(k) \\ \mathbf{y}(k) &= \mathbf{c}\mathbf{x}(k) \end{aligned} \quad (1a)$$

where  $\mathbf{x}(k)$  is an  $n \times 1$  state-variable vector,  $u(k)$  is a single input,  $\mathbf{y}(k)$  is a single output, and

$$\begin{aligned} \mathbf{A} &= \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \cdots & a_{nn} \end{bmatrix}, \quad \mathbf{b} = \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix} \\ \mathbf{c} &= [c_1 \quad c_2 \quad \cdots \quad c_n]. \end{aligned} \quad (1b)$$

The transfer function of the filter in (1a) can be expressed as

$$H(z) = \mathbf{c}(z\mathbf{I}_n - \mathbf{A})^{-1}\mathbf{b}. \quad (2)$$

The  $l_2$ -sensitivity measure for the filter in (1a) is defined as (Yan and Moore, 1992)

$$\begin{aligned} S &= \sum_{k=1}^n \sum_{l=1}^n \frac{1}{2\pi j} \oint_{|z|=1} \left| \frac{\partial H(z)}{\partial a_{kl}} \right|^2 \frac{dz}{z} \\ &+ \sum_{k=1}^n \frac{1}{2\pi j} \oint_{|z|=1} \left| \frac{\partial H(z)}{\partial b_k} \right|^2 \frac{dz}{z} \\ &+ \sum_{l=1}^n \frac{1}{2\pi j} \oint_{|z|=1} \left| \frac{\partial H(z)}{\partial c_l} \right|^2 \frac{dz}{z} \end{aligned} \quad (3a)$$

where

$$\begin{aligned} \frac{\partial H(z)}{\partial a_{kl}} &= \mathbf{G}(z)\mathbf{e}_k\mathbf{e}_l^T\mathbf{F}(z) \\ \frac{\partial H(z)}{\partial b_k} &= \mathbf{G}(z)\mathbf{e}_k, \quad \frac{\partial H(z)}{\partial c_l} = \mathbf{e}_l^T\mathbf{F}(z) \end{aligned} \quad (3b)$$

with

$$\mathbf{F}(z) = (z\mathbf{I}_n - \mathbf{A})^{-1}\mathbf{b}, \quad \mathbf{G}(z) = \mathbf{c}(z\mathbf{I}_n - \mathbf{A})^{-1}. \quad (3c)$$

It is noted that coefficients 0 and  $\pm 1$  can be realized precisely in the implementation of FWL digital systems. Therefore, system's  $l_2$ -sensitivity is not affected by these coefficients.

By taking this situation into account, the individual sensitivities for the elements of coefficient matrices  $\mathbf{A}$ ,  $\mathbf{b}$  and  $\mathbf{c}$  should be changed to (Xiao, 1997)

$$\begin{aligned} \frac{\partial H(z)}{\partial a_{kl}} &= \mathbf{G}(z)\mathbf{e}_k\mathbf{e}_l^T\mathbf{F}(z)\phi_{kl} \\ \frac{\partial H(z)}{\partial b_k} &= \mathbf{G}(z)\mathbf{e}_k\phi_k, \quad \frac{\partial H(z)}{\partial c_l} = \mathbf{e}_l^T\mathbf{F}(z)\psi_l \end{aligned} \quad (4a)$$

where

$$\begin{aligned} \phi_{kl} &= \begin{cases} 1 & \text{for } a_{kl} \neq 0, \pm 1 \\ 0 & \text{for } a_{kl} = 0, \pm 1 \end{cases} \\ \phi_k &= \begin{cases} 1 & \text{for } b_k \neq 0, \pm 1 \\ 0 & \text{for } b_k = 0, \pm 1 \end{cases} \\ \psi_l &= \begin{cases} 1 & \text{for } c_l \neq 0, \pm 1 \\ 0 & \text{for } c_l = 0, \pm 1 \end{cases}. \end{aligned} \quad (4b)$$

*Lemma:* The improved  $l_2$ -sensitivity measure for a state-space model  $(\mathbf{A}, \mathbf{b}, \mathbf{c})_n$  in (1a) is presented by (Xiao, 1997)

$$\begin{aligned} S_I &= \sum_{k=1}^n \sum_{l=1}^n \phi_{kl} [\mathbf{c} \quad \mathbf{0}] \mathbf{R}(k, l) \begin{bmatrix} \mathbf{c}^T \\ \mathbf{0} \end{bmatrix} \\ &+ \sum_{k=1}^n \phi_k W_{kk} + \sum_{l=1}^n \psi_l K_{ll} \end{aligned} \quad (5a)$$

where  $\mathbf{R}(k, l)$ ,  $K_{ll}$  ( $(l, l)$ th entry of  $\mathbf{K}$ ) and  $W_{kk}$  ( $(k, k)$ th entry of  $\mathbf{W}$ ) are obtained by solving the Lyapunov equations

$$\begin{aligned} \mathbf{R}(k, l) &= \begin{bmatrix} \mathbf{A} & \mathbf{e}_k\mathbf{e}_l^T \\ \mathbf{0} & \mathbf{A} \end{bmatrix} \mathbf{R}(k, l) \begin{bmatrix} \mathbf{A} & \mathbf{e}_k\mathbf{e}_l^T \\ \mathbf{0} & \mathbf{A} \end{bmatrix}^T \\ &+ \begin{bmatrix} \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{b}\mathbf{b}^T \end{bmatrix} \\ \mathbf{K} &= \mathbf{A}\mathbf{K}\mathbf{A}^T + \mathbf{b}\mathbf{b}^T \\ \mathbf{W} &= \mathbf{A}^T\mathbf{W}\mathbf{A} + \mathbf{c}^T\mathbf{c} \end{aligned} \quad (5b)$$

for  $k = 1, 2, \dots, n$  and  $l = 1, 2, \dots, n$ .

In the following two theorems, it is shown that the above improved  $l_2$ -sensitivity measure in (5a) can be modified to two novel forms so that the number of the Lyapunov equations is reduced.

*Theorem 1:* The improved  $l_2$ -sensitivity measure in (5a) is changed to the form

$$S'_l = \sum_{k=1}^n \sum_{l=1}^n \phi_{kl} \begin{bmatrix} e_l^T & \mathbf{0} \end{bmatrix} \mathbf{M}(k) \begin{bmatrix} e_l \\ \mathbf{0} \end{bmatrix} + \sum_{k=1}^n \phi_k W_{kk} + \sum_{l=1}^n \psi_l K_{ll} \quad (6a)$$

where  $\mathbf{M}(k)$  is obtained by solving the Lyapunov equation

$$\mathbf{M}(k) = \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix} \mathbf{M}(k) \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix}^T + \begin{bmatrix} \mathbf{0} & \mathbf{0} \\ \mathbf{0} & e_k e_k^T \end{bmatrix} \quad (6b)$$

for  $k = 1, 2, \dots, n$ .

*Proof:* It is noted that

$$\begin{aligned} \mathbf{F}(z)\mathbf{G}(z) &= (z\mathbf{I}_n - \mathbf{A})^{-1} \mathbf{bc}(z\mathbf{I}_n - \mathbf{A})^{-1} \\ &= \begin{bmatrix} \mathbf{I}_n & \mathbf{0} \end{bmatrix} \left( z\mathbf{I}_{2n} - \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix} \right)^{-1} \begin{bmatrix} \mathbf{0} \\ \mathbf{I}_n \end{bmatrix}. \end{aligned} \quad (7)$$

Defining  $\Phi(z) = \mathbf{F}(z)\mathbf{G}(z)$ , from (4a) and (7) it follows that

$$\begin{aligned} & \frac{1}{2\pi j} \oint_{|z|=1} \left| \frac{\partial H(z)}{\partial a_{kl}} \right|^2 \frac{dz}{z} \\ &= \phi_{kl} e_l^T \left[ \frac{1}{2\pi j} \oint_{|z|=1} \Phi(z) e_k e_k^T \Phi^T(z^{-1}) \frac{dz}{z} \right] e_l \\ &= \phi_{kl} \begin{bmatrix} e_l^T & \mathbf{0} \end{bmatrix} \mathbf{M}(k) \begin{bmatrix} e_l \\ \mathbf{0} \end{bmatrix} \end{aligned} \quad (8a)$$

where

$$\mathbf{M}(k) = \sum_{p=0}^{\infty} \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix}^p \begin{bmatrix} \mathbf{0} \\ e_k \end{bmatrix} \begin{bmatrix} e_k^T \\ \mathbf{0} \end{bmatrix} \begin{bmatrix} \mathbf{A}^T & \mathbf{0} \\ (\mathbf{bc})^T & \mathbf{A}^T \end{bmatrix}^p \quad (8b)$$

which yields the Lyapunov equation in (6b). This completes the proof of Theorem 1.

*Theorem 2:* The improved  $l_2$ -sensitivity measure in (5a) can be written as

$$S''_l = \sum_{k=1}^n \sum_{l=1}^n \phi_{kl} \begin{bmatrix} \mathbf{0} & e_k^T \end{bmatrix} \mathbf{N}(l) \begin{bmatrix} \mathbf{0} \\ e_k \end{bmatrix} + \sum_{k=1}^n \phi_k W_{kk} + \sum_{l=1}^n \psi_l K_{ll} \quad (9a)$$

where  $\mathbf{N}(l)$  is obtained by solving the Lyapunov equation

tion

$$\mathbf{N}(l) = \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix}^T \mathbf{N}(l) \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix} + \begin{bmatrix} e_l e_l^T & \mathbf{0} \\ \mathbf{0} & \mathbf{0} \end{bmatrix} \quad (9b)$$

for  $l = 1, 2, \dots, n$ .

*Proof:* From (4a) and (7) it follows that

$$\begin{aligned} & \frac{1}{2\pi j} \oint_{|z|=1} \left| \frac{\partial H(z)}{\partial a_{kl}} \right|^2 \frac{dz}{z} \\ &= \phi_{kl} e_k^T \left[ \frac{1}{2\pi j} \oint_{|z|=1} \Phi^T(z^{-1}) e_l e_l^T \Phi(z) \frac{dz}{z} \right] e_k \\ &= \phi_{kl} \begin{bmatrix} \mathbf{0} & e_k^T \end{bmatrix} \mathbf{N}(l) \begin{bmatrix} \mathbf{0} \\ e_k \end{bmatrix} \end{aligned} \quad (10a)$$

where

$$\mathbf{N}(l) = \sum_{p=0}^{\infty} \begin{bmatrix} \mathbf{A}^T & \mathbf{0} \\ (\mathbf{bc})^T & \mathbf{A}^T \end{bmatrix}^p \begin{bmatrix} e_l \\ \mathbf{0} \end{bmatrix} \begin{bmatrix} e_l^T \\ \mathbf{0} \end{bmatrix} \begin{bmatrix} \mathbf{A} & \mathbf{bc} \\ \mathbf{0} & \mathbf{A} \end{bmatrix}^p \quad (10b)$$

which yields the Lyapunov equation in (9b). This completes the proof of Theorem 2.

It is noted that the novel contribution in this paper exists in two alternative formulations in (6b) and (9b) for the Lyapunov equations where two independent variables ( $k, l$ ) in (5b) are replaced by a single independent variable either  $k$  or  $l$ . This makes it possible to reduce the amount of computations considerably.

### 3 MORE PRECISE $l_2$ -SENSITIVITY FOR A CLASS OF 2-D STATE-SPACE DIGITAL FILTERS

Consider a stable, locally controllable and locally observable 2-D local state-space model  $(\mathbf{A}_1, \mathbf{A}_2, \mathbf{b}, \mathbf{c}_1, \mathbf{c}_2, d)_n$  defined for a class of 2-D IIR digital filters (Hinamoto, 2001)

$$\begin{bmatrix} \mathbf{x}(i+1, j+1) \\ y(i, j) \end{bmatrix} = \begin{bmatrix} \mathbf{A}_1 & \mathbf{A}_2 \\ \mathbf{c}_1 & \mathbf{c}_2 \end{bmatrix} \begin{bmatrix} \mathbf{x}(i, j+1) \\ \mathbf{x}(i+1, j) \end{bmatrix} + \begin{bmatrix} \mathbf{b} \\ d \end{bmatrix} u(i, j) \quad (11a)$$

where  $\mathbf{x}(i, j)$  is an  $n \times 1$  local state vector,  $u(i, j)$  is a single input,  $y(i, j)$  is a single output, and

$$\begin{aligned} \mathbf{A}_i &= \begin{bmatrix} a_{i11} & a_{i12} & \cdots & a_{i1n} \\ a_{i21} & a_{i22} & \cdots & a_{i2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{in1} & a_{in2} & \cdots & a_{inn} \end{bmatrix}, & \mathbf{b} &= \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix} \\ \mathbf{c}_i &= [c_{i1} \quad c_{i2} \quad \cdots \quad c_{in}] & & \text{for } i = 1, 2. \end{aligned} \quad (11b)$$

The transfer function of (11a) is given by (Hinamoto, 2001)

$$H(z_1, z_2) = (z_1^{-1}c_1 + z_2^{-1}c_2) \cdot (\mathbf{I}_n - z_1^{-1}\mathbf{A}_1 - z_2^{-1}\mathbf{A}_2)^{-1} \mathbf{b} + d. \quad (12)$$

A more precise (an improved)  $l_2$ -sensitivity measure for the local state-space model in (11a) can be defined as

$$\begin{aligned} M = & \sum_{k=1}^n \sum_{l=1}^n \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial a_{1kl}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ & + \sum_{k=1}^n \sum_{l=1}^n \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial a_{2kl}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ & + \sum_{k=1}^n \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial b_k} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ & + \sum_{l=1}^n \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial c_{1l}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ & + \sum_{l=1}^n \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial c_{2l}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \end{aligned} \quad (13a)$$

where

$$\begin{aligned} \frac{\partial H(z_1, z_2)}{\partial a_{ikl}} &= z_i^{-1} \mathbf{G}(z_1, z_2) \mathbf{e}_k \mathbf{e}_l^T \mathbf{F}(z_1, z_2) \phi_{ikl} \\ \frac{\partial H(z_1, z_2)}{\partial b_k} &= \mathbf{G}(z_1, z_2) \mathbf{e}_k \phi_k \\ \frac{\partial H(z_1, z_2)}{\partial c_{il}} &= z_i^{-1} \mathbf{e}_l^T \mathbf{F}(z_1, z_2) \psi_{il} \quad \text{for } i = 1, 2 \end{aligned} \quad (13b)$$

with

$$\begin{aligned} \mathbf{F}(z_1, z_2) &= (\mathbf{I}_n - z_1^{-1}\mathbf{A}_1 - z_2^{-1}\mathbf{A}_2)^{-1} \mathbf{b} \\ \mathbf{G}(z_1, z_2) &= (z_1^{-1}c_1 + z_2^{-1}c_2) \cdot (\mathbf{I}_n - z_1^{-1}\mathbf{A}_1 - z_2^{-1}\mathbf{A}_2)^{-1} \\ \phi_{ikl} &= \begin{cases} 1 & \text{for } a_{ikl} \neq 0, \pm 1 \\ 0 & \text{for } a_{ikl} = 0, \pm 1 \end{cases} \\ \phi_k &= \begin{cases} 1 & \text{for } b_k \neq 0, \pm 1 \\ 0 & \text{for } b_k = 0, \pm 1 \end{cases} \\ \psi_{il} &= \begin{cases} 1 & \text{for } c_{il} \neq 0, \pm 1 \\ 0 & \text{for } c_{il} = 0, \pm 1 \end{cases} \end{aligned} \quad (13c)$$

for  $i = 1, 2$ .

It is noted that unlike those in (Hinamoto et al., 2006), the individual sensitivities in (13b) are taken into account 0 and  $\pm 1$  elements in the 2-D local state-space model of (11a) to evaluate the  $l_2$ -sensitivity more precisely.

By substituting (13b) into (13a), a more precise  $l_2$ -sensitivity measure for a 2-D state-space digital filter in (11a) is derived.

*Theorem 3*: The more precise  $l_2$ -sensitivity measure for the 2-D filter in (11a) can be computed from either of

$$\begin{aligned} M_I = & \sum_{k=1}^n \sum_{l=1}^n \phi_{1kl} \mathbf{e}_k^T \mathbf{M}_l \mathbf{e}_k + \sum_{k=1}^n \sum_{l=1}^n \phi_{2kl} \mathbf{e}_k^T \mathbf{M}_l \mathbf{e}_k \\ & + \sum_{k=1}^n \phi_k W_{kk} + \sum_{l=1}^n \psi_{1l} K_{ll} + \sum_{l=1}^n \psi_{2l} K_{ll} \end{aligned} \quad (14a)$$

$$\begin{aligned} M'_I = & \sum_{k=1}^n \sum_{l=1}^n \phi_{1kl} \mathbf{e}_l^T \mathbf{N}_k \mathbf{e}_l + \sum_{k=1}^n \sum_{l=1}^n \phi_{2kl} \mathbf{e}_l^T \mathbf{N}_k \mathbf{e}_l \\ & + \sum_{k=1}^n \phi_k W_{kk} + \sum_{l=1}^n \psi_{1l} K_{ll} + \sum_{l=1}^n \psi_{2l} K_{ll} \end{aligned} \quad (14b)$$

*Proof*: It follows from (13b) and (13c) that

$$\begin{aligned} & \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial a_{ikl}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ &= \phi_{ikl} \mathbf{e}_k^T \mathbf{M}_l \mathbf{e}_k = \phi_{ikl} \mathbf{e}_l^T \mathbf{N}_k \mathbf{e}_l \\ & \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial b_k} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ &= \phi_k \mathbf{e}_k^T \mathbf{W} \mathbf{e}_k \\ & \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \left| \frac{\partial H(z_1, z_2)}{\partial c_{il}} \right|^2 \frac{dz_1 dz_2}{z_1 z_2} \\ &= \psi_{il} \mathbf{e}_l^T \mathbf{K} \mathbf{e}_l \quad \text{for } i = 1, 2 \end{aligned} \quad (15a)$$

where

$$\begin{aligned} \mathbf{M}_l &= \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} [\mathbf{F}(z_1, z_2) \mathbf{G}(z_1, z_2)]^T \mathbf{e}_l \\ & \cdot \mathbf{e}_l^T \mathbf{F}(z_1^{-1}, z_2^{-1}) \mathbf{G}(z_1^{-1}, z_2^{-1}) \frac{dz_1 dz_2}{z_1 z_2} \\ \mathbf{N}_k &= \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \mathbf{F}(z_1, z_2) \mathbf{G}(z_1, z_2) \mathbf{e}_k \\ & \cdot \mathbf{e}_k^T [\mathbf{F}(z_1^{-1}, z_2^{-1}) \mathbf{G}(z_1^{-1}, z_2^{-1})]^T \frac{dz_1 dz_2}{z_1 z_2} \\ \mathbf{K} &= \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \mathbf{F}(z_1, z_2) \mathbf{F}^T(z_1^{-1}, z_2^{-1}) \frac{dz_1 dz_2}{z_1 z_2} \\ \mathbf{W} &= \frac{1}{(2\pi j)^2} \oint_{|z_1|=1} \oint_{|z_2|=1} \mathbf{G}^T(z_1, z_2) \mathbf{G}(z_1^{-1}, z_2^{-1}) \frac{dz_1 dz_2}{z_1 z_2}. \end{aligned} \quad (15b)$$

Matrices  $\mathbf{M}_l$ ,  $\mathbf{N}_k$ ,  $\mathbf{K}$  and  $\mathbf{W}$  are the 2-D Gramians

which can be derived from

$$\begin{aligned} \mathbf{M}_l &= \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \mathbf{H}^T(i, j) \mathbf{e}_l \mathbf{e}_l^T \mathbf{H}(i, j) \\ \mathbf{N}_k &= \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \mathbf{H}(i, j) \mathbf{e}_k \mathbf{e}_k^T \mathbf{H}^T(i, j) \\ \mathbf{K} &= \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \mathbf{f}(i, j) \mathbf{f}^T(i, j) \\ \mathbf{W} &= \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \mathbf{g}^T(i, j) \mathbf{g}(i, j) \end{aligned} \quad (16a)$$

where

$$\begin{aligned} \mathbf{f}(i, j) &= \mathbf{A}^{(i, j)} \mathbf{b} \\ \mathbf{g}(i, j) &= \mathbf{c}_1 \mathbf{A}^{(i-1, j)} + \mathbf{c}_2 \mathbf{A}^{(i, j-1)} \\ \mathbf{A}^{(0,0)} &= \mathbf{I}_n, \quad \mathbf{A}^{(i, j)} = \mathbf{0} \quad \text{for } i < 0, j < 0 \\ \mathbf{A}^{(i, j)} &= \mathbf{A}_1 \mathbf{A}^{(i-1, j)} + \mathbf{A}_2 \mathbf{A}^{(i, j-1)} \\ &= \mathbf{A}^{(i-1, j)} \mathbf{A}_1 + \mathbf{A}^{(i, j-1)} \mathbf{A}_2 \quad \text{for } (i, j) > (0, 0) \\ \mathbf{H}(i, j) &= \sum_{(0,0) \leq (k,r) < (i,j)} \mathbf{f}(k, r) \mathbf{g}(i-k, j-r) \end{aligned} \quad (16b)$$

with the partial ordering for integer pairs  $(i, j)$  defined in (Roessor, 1975).

## 4 NUMERICAL EXAMPLES

*Example 1:* Let a state-space digital filter  $(\mathbf{A}, \mathbf{b}, \mathbf{c})_3$  in (1a) be specified by (Xiao, 1997)

$$\mathbf{A} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0.1732 & -1.0227 & 1.8155 \end{bmatrix}, \quad \mathbf{b} = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$\mathbf{c} = [0.1174 \quad -0.3818 \quad 0.2984]$$

From (6a) and (9a), the improved  $l_2$ -sensitivity was found to be

$$S'_l = 240.433072, \quad S''_l = 240.433072$$

which essentially coincide with  $S_l = 240.43$  in (Xiao, 1997). The optimal state-space digital filter with the minimum  $l_2$ -sensitivity can be constructed as (Yan and Moore, 1992), (Xiao, 1997)

$$\mathbf{A}^o = \begin{bmatrix} 0.6883 & -0.2234 & -0.0297 \\ -0.2234 & 0.5394 & -0.1394 \\ -0.0297 & -0.1394 & 0.5879 \end{bmatrix}$$

$$\mathbf{b}^o = \begin{bmatrix} 0.5183 \\ 0.1718 \\ 0.0158 \end{bmatrix}$$

$$\mathbf{c}^o = [0.5183 \quad 0.1718 \quad 0.0158]$$

For this optimal realization, from (6a) and (9a), the improved  $l_2$ -sensitivity was found to be

$$S'_l = 2.458368, \quad S''_l = 2.458368$$

which essentially coincide with  $S_l = 2.4579$  in (Xiao, 1997). In fact,  $S_l = 2.458368$  was derived from (5a).

*Example 2:* Consider a local state-space model in (11a) specified by (Hinamoto et al., 2006)

$$\mathbf{A}_1 = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -0.0041 & 0.0801 & -0.4246 & 1.0446 \end{bmatrix}$$

$$\mathbf{A}_2 = \begin{bmatrix} -0.2261 & 1.6143 & 0.1005 & -0.0072 \\ -0.4059 & 1.6104 & -0.6062 & 0.2458 \\ -0.3096 & 1.0234 & -0.4532 & 0.3867 \\ -0.1447 & 0.4387 & -0.3102 & 0.5629 \end{bmatrix}$$

$$\mathbf{b} = [0 \quad 0 \quad 0 \quad 1]^T$$

$$\mathbf{c}_1 = [-0.0145 \quad 0.0123 \quad 0.0205 \quad 0.0476]$$

$$\mathbf{c}_2 = [0.01190 \quad 0.0235 \quad -0.0064 \quad 0.0209]$$

In (16a), the Gramians  $\mathbf{M}_l$ ,  $\mathbf{N}_k$ ,  $\mathbf{K}$ , and  $\mathbf{W}$  were computed over  $(0, 0) \leq (i, j) \leq (100, 100)$ . From (14a) and (14b), the more precise  $l_2$ -sensitivity was found to be

$M_l = 1.426236 \times 10^5$ ,  $M'_l = 1.426236 \times 10^5$  which are smaller than the value of the  $l_2$ -sensitivity measure:  $1.579936 \times 10^5$  reported in (Hinamoto et al., 2006) since coefficients 0 and  $\pm 1$  were not taken into account in (Hinamoto et al., 2006).

The optimal 2-D state-space digital filter structure that minimizes the  $l_2$ -sensitivity subject to  $l_2$ -scaling constraints was constructed as (Hinamoto et al., 2006)

$$\mathbf{A}_1^o = \begin{bmatrix} 0.31822 & 0.36329 & -0.21491 & -0.14635 \\ -0.01438 & 0.13734 & 0.56514 & -0.07384 \\ -0.08182 & -0.07710 & 0.16782 & 0.18698 \\ -0.01344 & 0.07553 & -0.03754 & 0.42122 \end{bmatrix}$$

$$\mathbf{A}_2^o = \begin{bmatrix} 0.53774 & -0.04378 & 0.15039 & 0.21219 \\ 0.08849 & 0.38433 & 0.01571 & 0.00945 \\ -0.22483 & 0.36107 & 0.11917 & -0.07048 \\ -0.09396 & -0.04963 & 0.14221 & 0.45275 \end{bmatrix}$$

$$\mathbf{b}^o = \begin{bmatrix} -0.38108 \\ -0.36371 \\ -0.77857 \\ 0.53703 \end{bmatrix}$$

$$\mathbf{c}_1^o = [-0.19351 \quad -0.07547 \quad 0.06028 \quad -0.01237]$$

$$\mathbf{c}_2^o = [0.15022 \quad 0.38727 \quad 0.40379 \quad 0.99327]$$

For this optimal realization that does not contain 0 and  $\pm 1$  elements, from (14a) and (14b) the more precise  $l_2$ -sensitivity was found to be

$M_l = 372.778156$ ,  $M'_l = 372.778156$  which essentially coincide with the value of an  $l_2$ -sensitivity measure: 372.776303 in (Hinamoto et al., 2006).

## 5 CONCLUSIONS

This paper has developed a simplified method of computing an improved  $l_2$ -sensitivity measure for state-space digital filters by reducing the number of the Lyapunov equations. The simplified method has also been expanded into a class of 2-D state-space digital filters. First, a conventional improved  $l_2$ -sensitivity for state-space digital filters has been reviewed and its computation method has been simplified with two novel forms such that the number of the Lyapunov equations is reduced. Next, the resulting method has been applied to a class of 2-D state-space digital filters. This has been done more precisely by taking into account 0 and  $\pm 1$  elements in the filter. Finally, two numerical examples have been presented to explain the validity and effectiveness of simplified methods of computing more precise (improved)  $l_2$ -sensitivity measures for 1-D as well as a class of 2-D state-space digital filters.

The simplified method has also been investigated for computing a more precise  $l_2$ -sensitivity measure in 2-D state-space digital filters described by the Roessor model (Roessor, 1975) and the results will appear elsewhere.

## REFERENCES

- Fornasini, E. and Marchesini, G. (1978). Doubly-indexed dynamical systems: State-space models and structural properties. *Math Syst. Theory*, 12:59–72.
- Gevers, M. and Li, G. (1993). *Parameterizations in Control, Estimation and Filtering Problems: Accuracy Aspects*. Springer-Verlag.
- Hinamoto, T. (2001). A novel local state-space model for 2-d digital filters and its properties. *Proc. 2001 IEEE Int. Symp. Circuits Syst.*, 2:545–548.
- Hinamoto, T., Hamanaka, T., and Maekawa, S. (Sept. 1990). Synthesis of 2-d state-space digital filters with low sensitivity based on the fornasini-marchesini model. *IEEE Trans. Acoust., Speech, Signal Processing*, ASSP-38:1587–1594.
- Hinamoto, T., Iwata, K., O.I.Omoifo, Ohno, S., and Lu, W.-S. (July. 2006). Optimal synthesis of a class of 2-d digital filters with minimum  $l_2$ -sensitivity and no overflow oscillations. *IEICE Trans. Fundamentals*, E89-A:1987–1994.
- Hinamoto, T., Ohnishi, H., and Lu, W.-S. (Oct. 2005). Minimization of  $l_2$ -sensitivity for state-space digital filters subject to  $l_2$ -dynamic-range scaling constraints. *IEEE Trans. Circuits Syst. II*, 52:641–645.
- Hinamoto, T., Yokoyama, S., Inoue, T., Zeng, W., and Lu, W.-S. (Sept. 2002). Analysis and minimization of  $l_2$ -sensitivity for linear systems and two-dimensional 2-d state-space filters using general controllability and observability gramians. *IEEE Trans. Circuits Syst. I*, 49:1279–1289.
- Hinamoto, Y. and Doi, A. (2012). Analysis of  $l_2$ -sensitivity for canonical forms in 1-d and 2-d separable-denominator digital filters. *Proc. 2012 IEEE Int. Midwest Symp. Circuits Syst.*, pages 920–923.
- Iwatsuki, M., Kawamata, M., and Higuchi, T. (Jan. 1989). Statistical sensitivity and minimum sensitivity structures with fewer coefficients in discrete time linear systems. *IEEE Trans. Circuits Syst.*, 37:72–80.
- Kawamata, M., Lin, T., and Higuchi, T. (1987). Minimization of sensitivity of 2-d state-space digital filters and its relation to 2-d balanced realizations. *Proc. 1987 IEEE Int. Symp. Circuits Syst.*, pages 710–713.
- Li, G. (July 1997). On frequency weighted minimal  $l_2$  sensitivity of 2-d systems using fornasini-marchesini lss model. *IEEE Trans. Circuits Syst. I*, 44:642–646.
- Li, G. (Mar. 1998). Two-dimensional system optimal realizations with  $l_2$ -sensitivity minimization. *IEEE Trans. Signal Processing*, 46:809–813.
- Li, G., Anderson, B. D. O., Gevers, M., and Perkins, J. E. (May 1992). Optimal fwl design of state-space digital systems with weighted sensitivity minimization and sparseness consideration. *IEEE Trans. Circuits Syst. I*, 39:365–377.
- Li, G. and Gevers, M. (1992). Optimal synthetic fwl design of state-space digital filters. *Proc. ICASSP 1992, San Francisco, CA*, 4:429–432.
- Li, G. and Gevers, M. (Dec. 1990). Optimal finite precision implementation of a state-estimate feedback controller. *IEEE Trans. Circuits Syst.*, 37:1487–1498.
- Roessor, R. P. (Feb. 1975). A discrete state-space model for linear image processing. *IEEE Trans. Automat. Contr.*, AC-20:1–10.
- Thiele, L. (Jan. 1984). Design of sensitivity and round-off noise optimal state-space discrete systems. *Int. J. Circuit Theory Appl.*, 12:39–46.
- Thiele, L. (May 1986). On the sensitivity of linear state-space systems. *IEEE Trans. Circuits Syst.*, CAS-33:502–510.
- Xiao, C. (Apr. 1997). Improved  $l_2$ -sensitivity for state-space digital systems. *IEEE Trans. Signal Processing*, 45:837–840.
- Yamaki, S., Abe, M., and Kawamata, M. (2006). A novel approach to  $l_2$ -sensitivity minimization of digital filters subject to  $l_2$ -scaling constraints. *Proc. 2006 IEEE Int. Symp. Circuits Syst.*, pages 5219–5222.
- Yamaki, S., Abe, M., and Kawamata, M. (2007). A fast convergence algorithm for  $l_2$ -sensitivity minimization of 2-d separable-denominator state-space digital filters. *Proc. 2007 IEEE Int. Symp. Circuits Syst.*, pages 2722–2725.
- Yan, W.-Y. and Moore, J. B. (Aug. 1992). On  $l^2$ -sensitivity minimization of linear state-space systems. *IEEE Trans. Circuits Syst. I*, 39:641–648.