NONLINEAR SYSTEM IDENTIFICATION USING DISCRETE-TIME NEURAL NETWORKS WITH STABLE LEARNING ALGORITHM

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Abstract: This paper presents a stable neural sytem identification for nonlinear systems. An input output discrete time representation is considered. No a priori knowledge about the nonlinearities of the system is assumed. The proposed learning rule is the backpropagation algorithm under the condition that the learning rate belongs to a specified range defining the stability domain. Satisfying such condition, unstable phenomenon during the learning process is avoided. A Lyapunov analysis is made in order to extract the new updating formulation which contain a set of inequality constraints. In the constrained learning rate algorithm, the learning rate is updated at each iteration by an equation derived using the stability conditions. As a case study, identification of two discrete time systems is considered to demonstrate the effectiveness of the proposed algorithm. Simulation results concerning the considered systems are presented.

1 INTRODUCTION

The area of system identification has received significant attention over the past decades and now it is a fairly mature field with many powerful methods available at the disposal of control engineers. Online system identification methods to date are based on recursive methods such as least squares, for most systems that are expressed as linear in the parameters (LIP).

To overcome this LIP assumption, neural networks (NNs) are now employed for system identification since these networks learn complex mappings from a set of examples. Due to NN approximation properties as well as the inherent adaptation features of these networks, NN present a potentially appealing alternative to modeling of nonlinear systems.

Moreover, from a practical perspective, the massive parallelism and fast adaptability of NN

implementations provide additional incentives for further investigation.

Several approaches have been presented for system identification without using NN and using NN (Narendra and Parthasarathy, 1990) (Boskovic and Narendra 1995). Most of the developments are done in continuous time due to the simplicity of deriving adaptation schemes. To the contrary, very few results are available for the system identification in discrete time using NNs. However, most of the schemes for system identification using NN have been demonstrated through empirical studies, or convergence of the output error is shown under ideal conditions (Ching-Hang Lee and al, 2002).

Others (Sadegh, 1993) have shown the stability of the overall system or convergence of the output error using linearity in the parameters assumption. Both recurrent and dynamic NN, in which the NN has its own dynamics, have been used for system identification.

Korkobi T., Djemel M. and Chtourou M. (2008). NONLINEAR SYSTEM IDENTIFICATION USING DISCRETE-TIME NEURAL NETWORKS WITH STABLE LEARNING ALGORITHM. In Proceedings of the Fifth International Conference on Informatics in Control, Automation and Robotics - ICSO, pages 351-356 DOI: 10.5220/0001504403510356 Copyright © SciTePress Most identification schemes using either multilayer feedforward or recurrent NN include identifier structures which do not guarantee the boundedness of the identification error of the system under nonideal conditions even in the open-loop configuration.

Recent results show that neural network technique seems to be very effective to identify a broad category of complex nonlinear systems when complete model information cannot be obtained. Lyapunov approach can be used directly to obtain stable training algorithms for continuous-time neural networks (Ge, Hang, Lee, Zhang, 2001), Polycarpou, (Kosmatopoulos, Christodoulou, Ioannou, 1995) (Yu, Poznyak, Li, 2001). The stability of neural networks can be found in (Feng, Michel, 1999)and (Suykens, Vandewalle, De Moor, 1997). The stability of learning algorithms was discussed in (Jin, Gupta, 1999) and (Polycarpou, Ioannou 1992).

It is well known that normal identification algorithms are stable for ideal plants (Ioannou, Sun, 2004). In the presence of disturbances or unmodeled dynamics, these adaptive procedures can go to instability easily. The lack of robustness in parameters identification was demonstrated in (E. Barn, 1992) and became a hot issue in 1980s. Several robust modification techniques were proposed in (Ioannou, Sun, 2004). The weight adjusting algorithms of neural networks is a type of parameters identification, the normal gradient algorithm is stable when neural network model can match the nonlinear plant exactly (Polycarpou, Ioannou 1992). Generally, some modifications to the normal gradient algorithm or backpropagation should be applied, such that the learning process is stable. For example, in (L. Jin, M.M. Gupta, 1999) some hard restrictions were added in the learning law, in (Polycarpou, Ioannou 1992) the dynamic backpropagation was modified with stability constraints.

The paper is organized as follows. Section II describes the adopted identification scheme. In section III and through a stability analysis a constrained learning rate algorithm is proposed to provide stable adaptive updating process. two simple simulation examples give the effectiveness of the suggested algorithm in section VI.

2 PRELIMINARIES

The main concern of this section is to introduce the feedfarward neural network adopted architechture

and some concepts of backpropagation training algorithm. Consider the following discrete-time input-output nonlinear system

$$y(k+1) = f[y(k)....y(k-n+1),u(k)...u(k-m+1)]$$
(1)

The neural model for the plant can be expressed as

$$\hat{y}(k+1) = F[Y(k), W, V]$$
(2)

Where
$$Y(k) = (y(k), y(k-1), \dots, y(k-n+1), u(k))$$

, $u(k-1), \dots, u(k-m+1)$

and W and V is the weight parameter vector for the neural model.

A typical multilayer feedfarward neural network is shown in *Figure 1*, where I_i is the ith neuron input, O_j is the jth neuron output i, j and k indicate neurons, wij is the weight between neuron i and neuron *j*. For the ith neuron, the nonlinear active function is defined as

$$f(x) = \frac{1}{1 + e^{-x}}$$
(3)

The output y^m (k) of the considered NN is

Wi

$$I_{j} = \sum_{i=1}^{n+m+1} x_{i}w_{ij} ; O_{j} = f(I_{j}); j=1,...,m$$

$$I' = \sum_{j=1}^{N} O_{j}V_{j} ; y^{m}(k) = f(I'); k=1,...,N$$

$$(4)$$



Figure 1: Feedforward neural model.

Training the neural model consists on the adjustment the weight parameters so that the neural model emulates the nonlinear plant dynamics. Input-output examples are obtained from the operation history of the plant.

Using the gradient decent, the weight connecting neuron i to neuron j is updated as

$$\begin{cases} W_{ij}(k+1) = W_{ij}(k) - \varepsilon \cdot \frac{\partial J(k)}{\partial W_{ij}(k)} \\ V_j(k+1) = V_j(k) - \varepsilon \cdot \frac{\partial J(k)}{\partial V_j(k)} \end{cases}$$
(5)

Where $J(k) = \frac{1}{2} [y(k+1) - y^m(k+1)]^2$

 ε is the learning rate. The partial derivatives are calculated with respect to the vector of weights *W*.

$$\begin{cases} \frac{\partial J(k)}{\partial V_j(k)} = f'(I_j)(y(k+1) - y^m(k+1))O_j \\ \frac{\partial J(k)}{\partial w_{ij}(k)} = f'(I_j) \left[\sum_{j=1}^{L} f'(I_j)(y(k+1) - y^m(k+1))V_j \right] x_i \end{cases}$$
(6)

Backpropagation algorithm has become the most popular one for training of the multilayer perceptron . Generally, some modifications to the normal gradient algorithm or backpropagation should be applied, such that the learning process is stable. For example, in (B. Egardt, 1979) some hard restrictions were added in the learning law, in (J.A.K. Suykens, J. Vandewalle, B. De Moor, 1997) the dynamic backpropagation was modified with stability constraints.

3 STABILITY ANALYSIS

In the literature, the Lyapunov synthesis (Z.P. Jiang, Y. Wang, 2001), (W. Yu, X. Li, 2001) consists on the selection of a positive function candidate V which lead to the computation of an adaptation law ensuring it's decrescence, i.e $\dot{V} \leq 0$ for continuous systems and $\Delta V(k) = V(k+1) - V(k) \leq 0$ for discrete time systems. Under these assumptions the function V is called Lyapunov function and garantee the stability of the system. Our objective is the determination of a stabilizing adaptation law ensuring the stability of the identification scheme presented below and the boundness of the output signals. The following assumptions are made for system (1)

Assumption 1. The unknown nonlinear function $f(\cdot)$ is continuous and differentiable.

Assumption 2. System output y(k) can be measured and its initial values are assumed to remain in a compact set Ω_0 .

3.1 Theorem

The stability of the identification scheme is guaranteed for a learning rate verifying the following inequality :

$$0 \le \varepsilon \le \frac{2 \left[tr \left(\frac{\partial J}{\partial W(k)} W^{T}(k) + \frac{\partial J}{\partial V(k)} V^{T}(k) \right) \right]}{\sum_{i,j} \left(\frac{\partial J}{\partial W_{ij}(k)} \right)^{2} + \sum_{ij} \left(\frac{\partial J}{\partial V_{j}(k)} \right)^{2}}$$
(7)

Where W, V are respectively the vector weight between the inputs and the hidden layer and the vector weight between the hidden layer and the outputs layer. i denote the ith input and j the jth hidden neuron.

3.2 Proof

Considering the Lyapunov function:

$$V_{L}(k) = tr\left(\widetilde{W}^{T}(k)\widetilde{W}(k)\right) + \widetilde{V}^{T}(k)\widetilde{V}(k)$$
(8)

Where

tr(.) denotes the matrix trace operation. $\tilde{V}(k) = V(k) - V^*$

$$\widetilde{W}(k) = W(k) - W^*$$

 W^* denotes the optimal vector weight between the inputs and the hidden layer.

 V^* denotes the optimal vector weight between the hidden layer and the outputs.

The computation of the $\Delta V_L(k)$ expression leads to :

$$\Delta V_{L}(k) = V(k+1) - V(k) = \left[tr\left(\widetilde{W}^{T}(k+1)\widetilde{W}(k+1)\right) + \widetilde{V}^{T}(k+1)\widetilde{V}(k+1)\right] - \left[tr\left(\widetilde{W}^{T}(k)\widetilde{W}(k)\right) + \widetilde{V}^{T}(k)\widetilde{V}(k)\right]$$
(9)

The adopted adaptation law is the gradient algorithm. We have:

$$\begin{cases} \widetilde{W}(k+1) = W(k) - \varepsilon \frac{\partial J}{\partial W(k)} - W^* \\ \widetilde{V}(k+1) = V(k) - \varepsilon \frac{\partial J}{\partial V(k)} - V^* \end{cases}$$

Where the partial derivatives are expressed as

$$\begin{cases} \frac{\partial J}{\partial W(k)} = \frac{\partial J}{\partial y(k+1)} \frac{\partial y(k+1)}{\partial W(k)} \\ \frac{\partial J}{\partial V(k)} = \frac{\partial J}{\partial y(k+1)} \frac{\partial y(k+1)}{\partial V(k)} \end{cases}$$
(10)

Our field of interest covers the black box systems. The partial derivatives denoting the system dynamic are approximated as follow:

$$\varepsilon_{s} = \frac{2\left[tr\left(\left\{\left[y^{m}(k+1)(1-y^{m}(k+1)).e(k)\sum_{j=1}^{N}V_{j1}.O_{j.}(1-O_{j})\right]x_{i}\right\}_{j\in[1\cdots m]}^{i\in[1\cdots m]} \cdot W^{T}(k) + \left\{\left[y^{m}(k+1)(1-y^{m}(k+1)).e(k).O_{j}\right]\right\}_{j\in[1\cdots m]} \cdot V^{T}(k)\right\}_{j\in[1\cdots m]} \cdot V^{T}(k) + \left[\left[y^{m}(k+1)(1-y^{m}(k+1)).e(k).O_{j}\right]\right]_{j\in[1\cdots m]} \cdot V^{T}(k)$$

$$\begin{cases} \frac{\partial y(k+1)}{\partial W(k)} \approx \frac{\partial y^m(k+1)}{\partial W(k)} \\ \frac{\partial y(k+1)}{\partial V(k)} \approx \frac{\partial y^m(k+1)}{\partial V(k)} \end{cases}$$
(11)

The approximated partial derivatives are given through:

$$\begin{bmatrix}
\frac{\partial J}{\partial W_{ij}(k)} \\
\begin{bmatrix}
\frac{\partial J}{\partial V_{j}(k)}
\end{bmatrix} = \begin{bmatrix}
y^{m}(k+1)(1-y^{m}(k+1))e(k)\sum_{j=1}^{N}V(j,l)O_{j}.(1-O_{j})\end{bmatrix}x_{i} \\
\begin{bmatrix}
\frac{\partial J}{\partial V_{j}(k)}
\end{bmatrix} = \begin{bmatrix}
y^{m}(k+1)(1-y^{m}(k+1))e(k).O_{j}\end{bmatrix}$$
(12)

Adopting the variables *A* and *B* defined by:

$$A = tr\left(\widetilde{W}^{T}(k+1)\widetilde{W}(k+1)\right) - tr\left(\widetilde{W}^{T}(k)\widetilde{W}(k)\right)$$
$$B = \widetilde{V}^{T}(k+1)\widetilde{V}(k+1) - \widetilde{V}^{T}(k)\widetilde{V}(k)$$

The $\Delta V(k)$ expression is calculated as :

$$\begin{split} \Delta V(k) &= A + B \\ &= tr \left(\varepsilon^2 \left[\frac{\partial J}{\partial W^T(k)} \frac{\partial J}{\partial W(k)} \right] - 2 \varepsilon \frac{\partial J}{\partial W(k)} \widetilde{W}(k) \right) \\ &+ \left(\varepsilon^2 \left[\frac{\partial J}{\partial V^T(k)} \frac{\partial J}{\partial V(k)} \right] - 2 \varepsilon \frac{\partial J}{\partial V(k)} \widetilde{V}(k) \right) \\ &= \left(\varepsilon^2 \sum_{i,j} \left[\frac{\partial J}{\partial W_{ij}(k)} \right]^2 - 2 \varepsilon tr \left(\frac{\partial J}{\partial W(k)} \widetilde{W}^T(k) \right) \right) \\ &+ \left(\varepsilon^2 \sum_{j} \left[\frac{\partial J}{\partial V_{j}(k)} \right]^2 - 2 \varepsilon \frac{\partial J}{\partial V(k)} \widetilde{V}^T(k) \right) \\ &= \varepsilon^2 \left(\sum_{i,j} \left[\frac{\partial J}{\partial W_{ij}(k)} \right]^2 + \sum_{j} \left[\frac{\partial J}{\partial V_{j}(k)} \right]^2 \right) \\ &- 2 \varepsilon \left(\frac{\partial J}{\partial V(k)} \widetilde{V}^T(k) + tr \left(\frac{\partial J}{\partial W(k)} \widetilde{W}^T(k) \right) \right) \\ &\leq \varepsilon^2 \left(\sum_{i,j} \left[\frac{\partial J}{\partial W_{ij}(k)} \right]^2 + \sum_{j} \left[\frac{\partial J}{\partial V_{j}(k)} \right]^2 \right) \\ &- 2 \varepsilon \left(\frac{\partial J}{\partial V(k)} V^T(k) + tr \left(\frac{\partial J}{\partial W(k)} W^T(k) \right) \right) \\ &\leq \alpha \cdot \varepsilon^2 - 2 \cdot \beta \cdot \varepsilon \end{split}$$

Where

$$\alpha = \sum_{i,j} \left[\frac{\partial J}{\partial W_{ij}(k)} \right]^2 + \sum_j \left[\frac{\partial J}{\partial V_j(k)} \right]^2$$
$$\beta = \frac{\partial J}{\partial V(k)} V^T(k) + tr \left(\frac{\partial J}{\partial W(k)} W^T(k) \right)$$

The stability condition $\Delta V(k) \le 0$ is satisfied only if :

$$\alpha \cdot \varepsilon^2 - 2 \cdot \beta \cdot \varepsilon \le 0 \tag{13}$$

Solving this ε second degree equation lead to the establishment of the condition (7) :

 $\Delta V(k) \le 0$ if ε satisfies the following condition : $0 \le \varepsilon \le \varepsilon_s$

where



Figure 2: Evolution of the system output and the neural model output ($\varepsilon \in stability \ domain$).

4 SIMULATION RESULTS

In this section two discrete time systems are considered to demonstrate the effectivness of the result discussed below.

4.1 First Order System

The considered system is a famous one in the litterature of adaptive neural control and identification. The discrete input-output equation is defined by:

$$y(k+1) = \frac{y(k)}{1+y(k)^2} + u(k)^3$$
(15)

For the neural model, a three-layer NN was selected with two inputs, three hidden and one output nodes. Sigmoidal activation functions were employed in all the nodes.

The weights are initialized to small random values. The learning rate is evaluated at each iteration through (14). It is also recognized that the training performs very well when the learning rate is small. As input signal, a sinusoidal one is chosen which the expression is defined by:

$$u(k) = 0.5 \cdot \cos\left(0.05k\pi + \frac{\pi}{5}\right)$$
 (16)

The simulations are realized in the two cases during 120 iterations. Two learnning rates values are fixed in and out of the learning rate range presented in (7).Simulation results are given through the following figures :



Figure 3 : Evolution of the system output and the neural model output ($\varepsilon \notin stability \ domain$).

4.1.1 Comments

Fig 2 and Fig 3 show that if the learning rate belongs to the range defined in (7), the stability of the identification scheme is garanteed. It is shown through this simulation that the identification objectives are satisfied. Out this variation domain of the learning rate, the identification is instable and the identification objectives are unreachable.

4.2 Second Order System

The second example concerns a discrete time system given by:

 $y(k) = 50 \tanh[\varphi(k-1)] + 0.5u(k-1)$ where $\varphi(k) = 2.10^{\circ} \left(\frac{(24+y(k))}{3} y(k-1) - 8 \frac{u^{2}(k)}{1+u^{2}(k)} y(k-1) \right)$ (17)

The process dynamic is interesting. In fact it has the behaviour of a first order law pass filter for inputs signal amplitude about 0.1, the behaviour of a linear second order system in the case of small amplitudes (0,1 < |u| < 0,5) and the behaviour of a non linear second order system in the case of great inputs amplitudes (0,5 < |u| < 5) (Ching-Hang Lee and al, 2002).

For the neural model, a three-layer NN was selected with three inputs, three hidden and one output nodes. Sigmoidal activation functions were employed in all the nodes.

The weights are initialized to small random values. The learning rate parameter is computed instantaneously. As input signal, a sinusoidal one is chosen which the expression is defined by:

$$u(k) = 0.5 \cdot \cos\left(0.005k\pi + \frac{\pi}{3}\right)$$
 (18)

The simulations are realized in the two cases. Two learning rates values are fixed in and out of the learning rate range presented in (7).

Simulation results are given through the following figures:



Figure 4: Evolution of the system output and the neural model output ($\varepsilon \in stability \ domain$).



Figure 5: Evolution of the system output and the neural model output ($\varepsilon \notin stability \ domain$).

4.2.1 Comments

Here we made a comparative study between an arbitrary choice of a learning rate out side of the stability domain and a constrained choice verifying the stability condition and guarantying tracking capability. The simulation results schow that a learning rate in the stability domain ensure the stability of the identification scheme.

5 CONCLUSIONS

To avoid unstable phenomenon during the learning process, constrained learning rate algorithm is proposed. A stable adaptive updating processes is guaranteed. A Lyapunov analysis is made in order to extract the new updating formulations which under inequality constraint. In the constrained learning rate algorithm, the learning rate is updated at each iterative instant by an equation derived using the stability conditions. The applicability of the approach presented is illustrated through two simulation examples.

REFERENCES

- B. Egardt, 1979. Stability of adaptive controllers. in: Lecture Notes in Control and Information Sciences 20, Springer-Verlag, Berlin.
- Z. Feng, A.N. Michel, 1999. Robustness analysis of a class of discrete-time systems with applications to neural networks, in: *Proceedings of American Control Conference*, San Deigo, pp. 3479–3483.
- S.S. Ge, C.C. Hang, T.H. Lee, T. Zhang, 2001: Stable Adaptive Neural Network Control, *Kluwer Academic*, Boston.

- P. A. Ioannou, J. Sun, 2004. Robust Adaptive Control, *Information Sciences* 158, 31–147, Prentice-Hall, Upper Saddle River.
- L. Jin, M.M. Gupta, 1999. Stable dynamic backpropagation learning in recurrent neural networks, *IEEE Trans. Neural Networks* 10 (6), 1321–1334.
- Z. P. Jiang, Y. Wang, 2001. Input-to-state stability for discrete-time nonlinear systems, *Automatica 37 (2)*, 857–869.
- E. B. Kosmatopoulos, M.M. Polycarpou, M.A. Christodoulou, P.A. Ioannou, 1995. High-order neural network structures for identification of dynamical systems, IEEE Trans. Neural Networks 6 (2), 431–442.
- M. M. Polycarpou, P.A. Ioannou 1992. Learning and convergence analysis of neural-type structured networks, *IEEE Trans. Neural Networks 3 (1)*, 39– 50.
- J. A. K. Suykens, J. Vandewalle, B. De Moor, 1997. NLq theory: checking and imposing stability of recurrent neural networks for nonlinear modelling, *IEEE Trans. Signal Process (special issue on neural networks for signal processing)* 45 (11), 2682–2691.
- W. Yu, X. Li, 2001. Some stability properties of dynamic neural networks, *IEEE Trans. Circuits Syst., Part I 48 (1)*, 256–259.
- W. Yu, A.S. Poznyak, X. Li, 2001. Multilayer dynamic neural networks for nonlinear system on-line identification, *Int. J. Control* 74 (18), 1858–1864.
- E. Barn, 1992. Optimisation for training neural nets, *IEEE Trans. Neural Networks 3 (2)*, 232–240.
- Ching-Hang Lee and al, 2002. Control of Nonlinear Dynamic Systems Via adaptive PID Control Scheme with Daturation Bound, *International Journal of Fuzzy Systems, Vol. 4 No. 4*,922-927.
- K. S. Narendra and K. Parthasarathy, 1990. Identification and control of dynamical systems using neural networks, *IEEE Transaction Neural Networks, vol.1, pp.4–27, Mar.*
- J. D. Boskovic and K.S.Narendra 1995. Comparison of linear nonlinear and neural-network based adaptive controllers for a class of fed-batch fermentation process, *Automatica*, vol. 31, no. 6, pp. 537-547.