

HMM-based Transient and Steady-state Current Signals Modeling for Electrical Appliances Identification

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Abstract: The electrical appliances identification problem is gaining a rapidly growing interest these past few years due to the recent need of this information in the new smart grid configuration. In this work, we propose to construct an appliance identification system based on the use of Hidden Markov Models (HMM) to model transient and steady-state electrical current signals. For this purpose, we investigate the usefulness of different choices for the proposed identification system such as: the use of the transient and the steady-state current signals, the use of even and odd-order harmonics as features, and the optimal number of features to take into account. This work also discusses the choice of the Short-Time Fourier Series (STFS) coefficients as adapted features for the representation of transient and steady-state current signals.

1 INTRODUCTION

The way power grids work to provide the needed electrical energy has changed a lot in the last few years. Classically, in the power grid, the energy goes only one way, i.e. from the power stations (usually thermal) to the consumers. With the advent of the idea of exploiting renewable energy resources (wind power, solar power, hydropower, etc.) the flow of energy in the power grid can no longer go only one way. A consumer having a wind turbine, solar cells, or other renewable-based energy generation systems becomes also an energy producer and this gives rise to a decentralized energy production. Also, in this new configuration, the energy production is no longer based on one energy source (thermal) but it can be generated using different, eventually renewable, energy sources. These new mutations created the need for an upgrade to the power grid and the result is what we call a smart grid.

As defined in (Gellings, 2009): “a smart grid is the use of sensors, communications, computational ability and control in some form to enhance the overall functionality of the electric power delivery system.” Adding these features to a power grid allows a bet-

ter monitoring and a continuous supervision of the energy flow. The energy sensors (or energy meters) along with their communication and computational capabilities are the basic building block of a smart grid. An energy meter allows the access to the energy consumption information of the appliance or group of appliances it measures.

The importance of energy metering and detailed electrical consumption information has been discussed in several previous works (Fischer, 2008) (Darby, 2006) (Darby, 2010) (Hancke et al., 2012). The impact of this information on the consumer behavior has been studied in (Chakravarty and Gupta, 2013) where the results showed that on average the consumers that used a solution that allowed the breakdown of the energy consumption consumed on average 14% less energy than the consumers that did not use it. According to (Carrie Armel et al., 2013), the benefits of appliance-level electrical consumption information over global consumption fall into three categories: (1) benefits to the consumer through possible energy savings thanks to the consumption feedback, (2) research and development benefits since the feedback allows the understanding of appliance energy consumption profiles and then redesigning ap-

pliances for energy efficiency, and (3) utility and policy benefits such as allowing for improved load forecasting (Feinberg and Genethliou, 2005) and energy demand prediction which is very important when it comes to decision making and taking economically reliable actions either by utilities or governments.

2 GENERAL BACKGROUND FOR ELECTRICAL APPLIANCES IDENTIFICATION

To break down the global energy consumption into its different constituent parts (for example, the whole home energy consumption into the appliance-level energy consumption) three approaches can be considered: intrusive, semi-intrusive, and non-intrusive approaches.

Intrusive Load Monitoring (ILM) approaches are based on the use of distributed sensing meters to support control, monitor and intervention of such devices (Burbano Acuña, 2015). As stated by (Parson, 2012) different forms of ILM can be found:

- Electrical sub-metering: one energy meter is used for each appliance
- Smart appliances: the appliances have communicating chips that allows them to self-report their energy consumption to a central unit
- Electrical probing: transmitting an electrical signal into the households mains circuit and analyzing the return signal

The ILM do not seem to be viable solutions since they assume the installation of different energy meters in different household locations which is not easy to do and most of the time costly. Semi-Intrusive Load Monitoring (SILM) is somehow a relaxed version of the ILM, where instead of using energy meters one per appliance, we use one energy meter to measure a group of appliances (Tang et al., 2014). SILM share the same drawbacks of the ILM but with less intrusion. Very little information can be found, in the literature, on SILM and this may be because they are usually confused with ILM approaches.

Non-Intrusive Load Monitoring (NILM) is a field where the main concern is to break down an aggregated energy consumption into its different constituent parts in a non-intrusive way using only one energy meter. For example, instead of introducing different energy meters put all around the household we only put one energy meter at the main breaker panel level. The work on this field started with Hart in the late 1980s (Hart, 1989) (Hart, 1992) where he

proposed to use the (real) power consumption variation to identify household appliances. Even though some other work was done in this field after Hart between the early 1990s and the late 2000s (Sultanem, 1991) (Leeb et al., 1995) (Cole and Albicki, 1998) (Drenker and Kader, 1999) (Chan et al., 2000) (Baranski and Voss, 2003) (Laughman et al., 2003) (Ting et al., 2005) (Patel et al., 2007) (Chang et al., 2008) it was not until around five years ago that this field started to gain a rapidly growing interest. A state of the art for the NILM techniques can be found in (Najmeddine et al., 2008) (Du et al., 2010) and (Zeifman and Roth, 2011). In (Carrie Armel et al., 2013) the authors gave a table that summarizes some of the works done in the NILM field up to 2012. They organized them in a chronological order and gave some details on the used methods, data types, appliance types, data frequency, and other interesting characteristics.

In this study we propose the use of Hidden Markov Models (HMM) for transient electrical signals modeling and classification. The obtained features are to be used eventually as a complementary information that helps a NILM system identify home appliances. Previous work on NILM using HMM was done by (Zia et al., 2011) (Beckel et al., 2012) (Parson, 2012) and more recently by (Ridi and Hennebert, 2014) and (Parson et al., 2014). In these studies, the authors applied their proposed approaches on low-frequency sampled signals (for example, (Beckel et al., 2012) used the REDD dataset¹ with 1 Hz sampling frequency signals and (Ridi and Hennebert, 2014) used the ACS-F1 dataset² that has 10 Hz sampling frequency signals). In our study, we propose the use of HMM on high-frequency sampled signals and we use for this purpose the recently released PLAID dataset (PLAID, 2015). It is worth mentioning that the lack of a *proper* database that is informative, diverse, and scalable (Lai et al., 2012) of high-frequency sampled electrical signals for appliances hindered a lot the development of high-frequency NILM. This was one of the concerns that have been discussed by the NILM community during the 2nd European NILM workshop (July, 2015) held in London³.

In the literature, and to the best of our knowledge, the authors of (Thiruvaran et al., 2013) are the only ones, to date, that have done a study on high-frequency sampled signals (10 kHz) using HMM. In (Thiruvaran et al., 2013) the authors proposed to

¹<http://redd.csail.mit.edu/>

²<http://www.wattict.com/web/index.php/databases/acs-f1>

³<http://www.nilmm.eu/>

use the Short-Time Fourier Transform (STFT) and Wavelet Transform (WT) coefficients as features for their identification system and as a dataset they collected measurements of current and voltage waveforms for four different appliances (a fluorescent lamp, an incandescent lamp, a computer monitor, and a motor). To acquire the waveforms, these four appliances were used to form switching-on sequences (24 possible sequence in total). For example, as a sequence we can imagine: the fluorescent lamp is first turned-on, then the incandescent lamp, then the computer monitor and then the motor. Finally, they announced accuracies of 97.9% when using the STFT features and 93.75% when using the WT features. In our study, we propose to consider not only the transient but also the steady-state and we quantify the usefulness of the transient part over the steady-state part for appliance recognition. Also, the accuracies obtained in (Thiruvaran et al., 2013) can not quantify the real recognition accuracy of the HMM-based system due to the small size of the used dataset. This is why the use of the PLAID dataset should allow more reliable results. Along with this, we propose in our study the evaluation of the harmonic-order on the recognition accuracy of the proposed HMM-based system instead of just taking into account all the harmonics. Another difference compared to the work done in (Thiruvaran et al., 2013) is that the PLAID dataset signals are electrical signals of individual appliances. That means that each signal was acquired with only one appliance operating alone. Analyzing such signals should allow the determination of *intrinsic* high-frequency characteristic features for the different appliances that can be used afterwards as a complementary information in a larger recognition system with data fusion capabilities.

3 HARMONIC ANALYSIS

There are different ways to analyze the harmonic contents of a signal. The most known method is the use of the Fourier Transform (FT). For a discrete-time signal $s[n]$ we define the Discrete Fourier Transform (DFT) as follows (Mallat, 1999):

$$S[k] = \sum_{n=0}^{L-1} s[n] \exp\left(\frac{-j2\pi kn}{L}\right), \quad k = 0, \dots, L-1 \quad (1)$$

where L is the length of $s[n]$ in samples. The computation of the harmonic content of $s[n]$ using Equation 1 starts to become heavier as L becomes bigger and bigger even when using fast algorithms like the Fast Fourier Transform (FFT), especially, when the

dataset is big and a lot of signals have to be analyzed. Since the electrical current signals are periodic (the steady-state signals), we propose to exploit this characteristic to improve the computational time using the Fourier Series (FS) decomposition instead of the FT. The FS are particular instances of the FT for Dirac sums (Mallat, 1999). This means that to get the FT of a periodic signal we only have to compute its FS coefficients. For the *periodic* signal $s[n]$, the Discrete FS (DFS) decomposition is:

$$s[n] = \sum_{k=0}^{N-1} C_k \exp\left(\frac{j2\pi kn}{N}\right), \quad (2)$$

where N is the period of $s[n]$ in samples and

$$C_k = \frac{1}{N} \sum_{n=0}^{N-1} s[n] \exp\left(\frac{-j2\pi kn}{N}\right), \quad k = 0, \dots, N-1 \quad (3)$$

are the DFS coefficients of $s[n]$. Even though, Equations 1 and 3 look similar (up to a factor), the main difference is the number of samples over which the sum is computed. By summing over the signal period N (usually $< L$) instead of over the whole signal length we gain in computational time.

The DFS coefficients C_k , $k = 1, \dots, N-1$ (Equation 3) correspond to the fundamental frequency of the periodic signal ($k = 1$) and its harmonics. For the steady-state electrical current signals these frequencies appear to represent most, if not all, of the information contained in the signal. Hence, this is another reason that justifies the use of the DFS instead of the DFT.

For transient electrical current signals, however, the periodicity property is lost and strictly speaking we are not allowed to use Equations 2 and 3. Nevertheless, we think that even for the transient signals, the important of the signal information is contained *around* the fundamental frequency and its harmonics. That is why we chose to represent the transient signals also using the DFS coefficients.

Since our approach is based on dividing the electrical current signals (transient and steady-state) into overlapping windows and computing the DFS on each window, the obtained coefficients can be called Short-Time Fourier Series (STFS) coefficients. Hence, for the remainder of the paper we will use the notation STFS instead of DFS. Each STFS coefficients' vector (computed on a window) is called an observation. Thus, each waveform (an electrical current signal) is represented by a sequence of vectors considered as a sequence of observations for the HMM-based classification.

4 EXPERIMENTAL RESULTS AND DISCUSSION

The appliance identification system is mainly based on three things: a dataset, a feature extraction (eventually feature selection) algorithm, and a classifier. As dataset we chose the PLAID dataset for which we give a brief description in the next sub-section. This dataset contains current and voltage waveforms (sometimes instantaneous power) of different appliances. For our study, we have decided to only use the current waveforms as an appliance signature for two main reasons. The first one is that the voltage waveform does not change much for the different appliances and hence does not add much information to the identification system. Actually, from a theoretical point of view, the voltage waveform is not supposed to change. We expect the voltage waveform to always be a sine wave with a fixed frequency equal to the power line frequency of 50 or 60 Hz (depending on the considered country) and a fixed root-mean-square (rms) voltage. Even though this is not a hundred percent true (since there are always fluctuations in the power grid characteristics) we can assume to a certain degree that it is the case in practice. The second reason is that we found on the dataset a lot of voltage waveforms that are incorrectly calibrated (i.e. having incorrect amplitude values that diverge a lot from the standard rms value of the grid voltage).

The used feature extraction algorithm is based on the STFS already discussed in section 3 and the considered features are the magnitudes of the STFS coefficients. For the classifier we chose to use Hidden Markov Models (HMM) for their capabilities to model dynamic time variations of signals.

An HMM is a finite state machine which changes state once every time unit (the value of the time unit depends on the observed phenomenon). Each time t that a state j is entered, an observation vector \mathbf{o}_t is generated from a probability density $b_j(\mathbf{o}_t)$. The transition from state i to state j is also probabilistic and is governed by the discrete probability a_{ij} . The emission likelihood b_j for state j and observation \mathbf{o}_t at time t is done by a Gaussian Mixture Model (GMM) (Hacine-Gharbi et al., 2012):

$$b_j(\mathbf{o}_t) = \sum_{m=1}^M c_m \mathcal{N}(\mathbf{o}_t; \boldsymbol{\mu}_m, \boldsymbol{\Sigma}_m), \quad (4)$$

where $\mathcal{N}(\mathbf{o}; \boldsymbol{\mu}, \boldsymbol{\Sigma})$ is the value in \mathbf{o} of a multivariate Gaussian with mean $\boldsymbol{\mu}$ and covariance $\boldsymbol{\Sigma}$. M Gaussians are used in a mixture, each weighed by c_m .

The classifier is based on modeling each appliance type by an HMM using the Hidden Markov Model

Toolkit (HTK) library (Young et al., 2009). The training is done in several steps by applying the embedded Baum–Welch reestimation (HEREST command). For the identification step we used the Viterbi algorithm (HVITE command). More details on the algorithms used for training and identification can be found in (Young et al., 2009). For our identification system, we chose to model each appliance type using 3 states, and each state using 3 Gaussian mixtures. Each waveform was subdivided into overlapping windows and from each window N features can be extracted (even though in practice we only choose a reduced number to work with, see the end of subsection 4.4). The window size was fixed to 16.7 msec (500 samples at 30 kHz frequency) that corresponds to the 60 Hz cycle-time and the overlapping to 50% of the window size.

The identification results are evaluated using the Classification Rate (CR) defined as:

$$CR(\%) = \frac{T - M}{T} \times 100, \quad (5)$$

where T is the total number of tested waveforms (each one representing an appliance) given to the input of the classifier and M is the number of misclassified tested waveforms.

4.1 Summary of the PLAID Dataset

The Plug Load Appliance Identification Dataset (PLAID) is a public dataset of electric signatures. These signatures are current and voltage measurements taken during the summer of 2013 in Pittsburgh, Pennsylvania, USA, from 55 households. This dataset contains 11 appliance types and for each appliance from three to six instances (the dataset contains a total of 1074 instances). Each signature of the dataset is a few-second-long signal containing the turn-on transient (when available, since for some appliances like the “Fridge” for example, which is usually working all the time, the turn-on transient is not always present) and a portion of the steady-state part (that corresponds to the steady consumption phase). These signals are sampled at a 30 kHz rate. Table 1 summarizes the appliances found on the dataset: the different appliance types and the number of instances for each type. For more details on the dataset please refer to (Gao et al., 2014). Finally, it is worth pointing out that the values of the number of instances we found in the dataset for each appliance type were most of the time different than the ones given in (Gao et al., 2014). The total number of instances in the dataset is also different. This might be due to an update of the dataset after the publication of the paper (Gao et al., 2014).

Table 1: Summary of the appliances found on the PLAID dataset.

Appliance type	Number of instances
Air Conditioner	66
Compact Fluorescent Lamp	175
Fan	115
Fridge	38
Hairdryer	156
Heater	35
Incandescent Light Bulb	114
Laptop	172
Microwave	139
Vacuum cleaner	38
Washing Machine	26
Total	1074

4.2 PLAID Dataset Subdivision for Training and Tests

In this work, the dataset is divided into a training set that allows the learning of appliances HMM models and a test set for the performance evaluation of the identification system. It is considered that all the houses (55 in total) have examples in the training and in the test sets. In order to study the effect of the number of waveform versions belonging to training and test sets, we proposed four subdivisions. In the first one, we take for each house, at most one version of the reference signal corresponding to a particular appliance and the other versions are kept for testing. This subdivision gives us 230 training waveforms and 844 test waveforms. The other subdivisions are obtained by changing the maximum number of waveform versions to put in the training set from 2 to 4. Thus, we got the four subdivisions shown in Table 2, where sb_i

Table 2: Different tested subdivisions of the dataset

sb #	sb1	sb2	sb3	sb4
NTR	230	441	648	848
NTEST	844	633	426	226

indicates the subdivision with “i” waveform versions in the training set. NTR is the number of the considered training waveforms and NTEST is the number of the considered test waveforms. For the following sub-sections, the results are given for each one of the above mentioned subdivisions.

4.3 Transient vs. Steady-state Electrical Signals Usefulness for Appliance Identification

The goal of the experiment done in this sub-section is to study the usefulness of the transient and the steady-state parts of the electrical current signals for the identification task of the different appliances. This experiment required us to segment the PLAID dataset waveforms each into a transient part and a steady-state part. This segmentation was done by characterizing the transient part as the one with the large energy variation compared to the steady-state part. For the performance evaluation, the classification rate CR using the transient part (TP), steady-state part (SSP) and the overall part (OP) was computed. This evaluation takes into account the different dataset subdivisions mentioned in sub-section 4.2. Note that in this experiment, we only used 10 STFS coefficients. Table 3 gives the CR for each subdivision sb_i each time considering a different waveform part (OP, TP and SSP).

Table 3: Transient vs. Steady-state CR evaluation.

sb #	sb1	sb2	sb3	sb4
CR (%), OP	86.02	87.36	87.56	88.50
CR (%), TP	88.98	91.15	90.85	92.48
CR (%), SSP	86.61	88.94	86.62	88.50

From these results, we can say that the use of the transient part is the one that gives the best CR . Moreover, the results show that the computational time when using only the transient part is much lower compared to the others. For example, in the case of sb_2 , the running test time was 218 seconds for TP, 418 seconds for SSP, and 436 seconds for OP. The hardware used for the experiment is a computer with an Intel Core i5 processor and 4 GB of RAM memory.

Thereby, the use of the transient should allow the construction of an appliance identification system that is fast and accurate which justifies its usefulness for the appliance identification task. In order to improve the identification, we studied further the harmonic analysis by taking into account the harmonic order (even or odd) and the optimal number of STFS coefficients to choose. The details are given in the next sub-sections.

4.4 Even vs. Odd Harmonic Order for Appliance Identification

The experiment presented in this sub-section allows the study of the harmonic order (even or odd order) relevance for the appliance identification task.

This experiment is similar to the one done in sub-section 4.3 but we only consider the transient part, following the results of sub-section 4.3. For the performance evaluation, we compared the identification system CR when using even harmonics only, odd harmonics only, and when using both. For each one of these configurations, we used the corresponding first 5 STFS coefficients as features (i.e, first 5 even harmonics, first 5 odd harmonics, and first 5 harmonics independently of the order, respectively). Table 4 gives the obtained CR for each tested configuration. From Table 4, we can deduce the following conclu-

Table 4: Even vs. Odd CR evaluation.

sb #	sb1	sb2	sb3	sb4
CR (%), (even+odd)	88.98	91.15	90.85	92.48
CR (%), even	63.63	64.45	64.79	63.27
CR (%), odd	90.05	89.57	89.44	90.71

sions:

- The odd-order harmonics give the best performance results no matter the considered subdivision w.r.t. the even-order harmonics
- The even-order harmonics give the worst performance results
- Adding the even-order harmonics to the odd-order harmonics feature vector does not seem to improve a lot the CR

To conclude, the coefficients of the fundamental and its odd harmonics are the most useful for the appliance identification task. This can be explained by the half-wave symmetry usually found in electrical signals (i.e. for a periodic signal $g(t)$ with period T , we have $g(t) = -g(t \pm T/2)$) (Nait Meziane et al., 2015). This half-wave symmetry causes the even order coefficients to have null values.

4.5 Optimal Number of Harmonics for Appliance Identification

A very important parameter to take into account is the optimal number (the smallest) of STFS coefficients to consider in order to guarantee the best CR . To get an insight on what this optimal number is, we evaluated the identification system using various feature vector (STFS coefficients number) sizes. We only considered the odd harmonics coefficients from 1 to 50. Figure 1 gives the obtained results. This figure shows the CR function of the odd harmonics number

taken into account for the identification. The results show that no significant improvement is noticed when taking more than 4 odd harmonics.

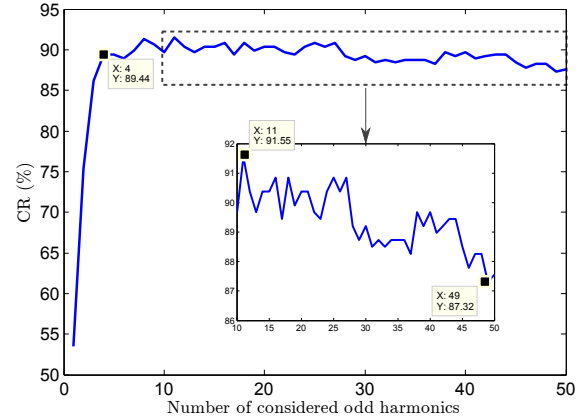


Figure 1: CR function of the considered odd harmonics number.

Moreover, after reaching a peak value (when taking 11 odd harmonics) and after fluctuating around a more or less stable value (when taking up to 27 odd harmonics) that is smaller than the peak value, the CR starts decreasing (see the zoom on Figure 1) as we start taking more and more coefficients. This phenomenon is very known in pattern recognition and is called the curse of dimensionality or peaking phenomenon (Jain et al., 2000). The dimension of the feature vector depends on the dimension of the dataset used. To avoid this problem, and as a rule of thumb, we usually say that the number of training data points should be an exponential function of the feature vector dimension (Jain et al., 2000).

5 CONCLUSIONS

In this paper, we have discussed the use of HMM models to solve the electrical appliance identification problem based on high-frequency sampled signals. We have evaluated the usefulness of different choices for the identification system (the use of transient vs. steady-state signals, even vs. odd-order harmonics as features, and the optimal feature vector size). We conclude from this study that the combined use of the transient part of the electrical current signals with only a few odd-order harmonics allows to construct an appliance identification system that is accurate, fast, and less complex in terms of memory occupancy and computations. In future work, and for more complete results, the use and comparison of different types of classifiers and different features may be considered.

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