

HIERARCHICAL AND PSEUDO-RANDOM EIRA CODES BASED ON BIBDS AND PRIMITIVE INTERLEAVERS

Jesús Miguel Pérez Llano, Víctor Fernández Solórzano

Microelectronics Engineering Group TEISA, ETSIIyT, Cantabria University, Av. Los Castros, Santander, Spain

Keywords: LDPC, eIRA, DVB-S2, BIBD, error floor.

Abstract: This paper describes a method based on hierarchical matrices and primitive generators that allows low cost coder and decoder implementations. The hierarchical approach is well suited for decoder implementation and, in addition, the method has been applied to eIRA structures which have demonstrated a reduced coder implementation complexity. Despite the added structure to eIRA original codes, the architecture presented shows similar BER performance. To achieve this, BIBDs have been used to avoid length-four cycles and primitive generators contribute to get a pseudo-random construction. Moreover, the reduction of low weight codewords and near codewords are considered in order to reduce error-floors.

1 INTRODUCTION

Although LDPC codes (Gallager, 1962)(Mackay, 1999) have been well known in research for a long time, they have reached the “standard world” through standards such as IEEE 802, some NASA standards and DVB in the recent DVB-S2 standard (DVB-S2, 2004). LDPC capabilities led to them being adopted for the previously mentioned standards: on the one hand because they are easier to decode than well known Turbo Code based systems and with more flexible architectures, and on the other hand because of their BER performance. Moreover, their implementation throughput makes them an efficient architecture-aware channel coding scheme.

In spite of all these advantages, LDPCs have some disadvantages from the point of view of hardware design whose effects are worth mitigating (if the BER performance is not greatly affected). Randomly designed LDPCs (Mackay, 1997) have been demonstrated to have superior BER performance. However, their implementation cost becomes unviable when the parity matrix size increases. Moreover, the codification of this kind of code generally becomes non sparse.

Some alternatives have been proposed to avoid the aforementioned implementation difficulties. Finite Geometry based LDPCs (Kou, 2001) have a greatly simplified codification but in general their matrices are regular and very dense, which makes their implementation cost a relevant drawback.

In (Yang, 2004) a new class of LDPCs called eIRAs was recently shown to have a simplified codification scheme, with good BER performance. Their biggest flaw is that a considerable part of its structure remains random (with the problems previously alluded to).

The implementation proposed in this paper is based on hierarchical matrices, which have been demonstrated, in (Mansour, 2003)(Mansour, 2004) by Mansour and in (Liao, 2004) by Liao and Yeo, to be suitable for decoder implementation. Mansour suggests the use of well known architectures such as Ramanujan (Rosenthal, 2000) or cyclotomic sets (Mansour, 2002) which have good structural properties but with some weaknesses, as the error floor due to low weight codewords (Mackay, 2003) and the lack of flexibility (Zhang, 2004). Liao and Yeo propose a completely random top and bottom architecture that do not ensure the absence of length-four cycles which has been demonstrated to have an important influence on BER. In contrast, the basis of our design is a well defined architecture based on a BIBD (Ammar, 2004)(Ammar, 2002) design at the top level architecture -which avoids length-four cycles- and pseudo random permutations with

This work has been supported by the Spanish MEC through the TEC2005-03301 project.

Miguel Pérez Llano J. and Fernández Solórzano V. (2006).
HIERARCHICAL AND PSEUDO-RANDOM EIRA CODES BASED ON BIBDS AND PRIMITIVE INTERLEAVERS.
In *Proceedings of the International Conference on Wireless Information Networks and Systems*, pages 151-157
Copyright © SciTePress

optimal roots at the bottom level designed to avoid low weight codewords and near codewords following recommendations of (Yang, 2004)(Dinoi, 2005)(Tian, 2004). Moreover, our proposal not only reduces the complexity of the decoder, as previously mentioned works do, but it simplifies the coder complexity.

The use of primitive generators (Morelos-Zaragoza, 2002) for the bottom level matrices enables the implementation presented to avoid storing the pointers, necessary to locate the 1's in the parity check matrix (the only pointers to be stored are those that define the BIBD structure which require much less memory consumption) without losing randomness.

The approach presented outperforms the congruent sequences based LDPCs (Prabhakar, 2002), which avoid the memory for the pointers but are more restrictive, being regular and only defining a bit degree of 3 in contrast to the architecture presented which is much more flexible.

The use of eIRA codes provides our design with a high performance from the point of view of BER. Moreover, the absence of cycles 4 combined with the optimal distribution of the connections in the parity matrix lead to minimize low weight codewords and near codewords, thus lowering the error floor. In Section II the architecture proposed will be discussed. In section III the BER performance of the scheme is presented. Finally, in chapter IV, the conclusions of this paper are presented.

2 ARCHITECTURE PROPOSED

LDPCs based on eIRAs (Yang, 2004) are constructed dividing the parity check matrix into two parts called H_1 and H_2 . H_2 is defined as a square matrix of size m (where m is the number of parity bits) composed of $m-1$ degree-2 columns and one degree-1 column (see Figure 1). In contrast, H_1 does not have a defined architecture, being designed at random (avoiding length-four cycles by removing appropriate bits) in (Yang, 2004).

In the architecture proposed H_1 is divided into several smaller matrices which are permutations of the identity matrix or square zero matrices (see Figure 1).

The TOP level architecture of H_1 will be chosen to avoid length-four cycles using the BIBD structure as will be explained in II-A. The sub-matrices, which are permutations of the identity matrix, are called permutation matrices and their structure and

construction will be shown in II-B. The absence of length-four cycles in TOP level architecture ensures a length-four cycle free parity check matrix because sub-matrices have no more than one '1' per column.

The architecture proposed is independent of the frame size and only depends on the bit degree and rate. To obtain different sizes the only thing that needs to be modified in the design is the size of bottom matrices, keeping the top matrix unchanged. Bearing in mind that positions to be stored are those of the top matrix, as code size increases, the architecture proposed needs relatively less memory than classical approaches.

2.1 TOP Level Architecture

The importance of avoiding cycles of length 4, for BER performance of LDPC based codes, has been widely demonstrated (Tian, 2004)(Mao, 2001). With this idea in mind, the design presented develops a way to design parity check matrices distributing the permutation matrices in such a way that they do not form length-four cycles. Moreover, this method can be used with various matrix sizes and rates.

The approach presented makes use of BIBDs (Balanced Incomplete Block Designs) (Anderson, 1990) in order to achieve a length-four cycle free matrix with minimum dimension for a given bit degree. BIBDs combine a number of points to form

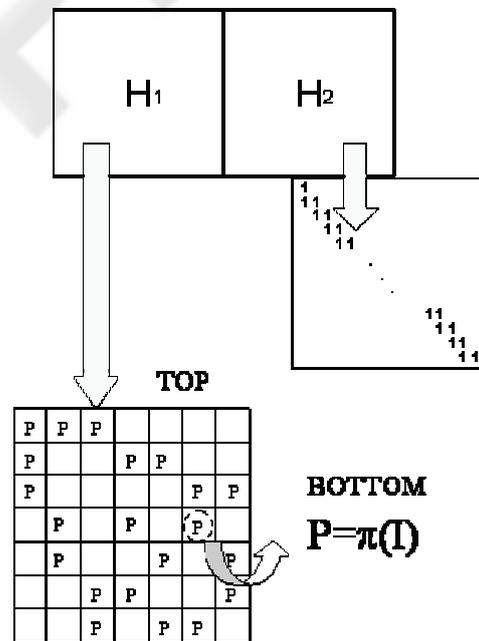


Figure 1: Proposed eIRA parity check matrix.

groups. They are defined by -at least- three parameters called v , r , λ , where v is the total number of points, r is the number of blocks which a point is in (i.e. the bit degree in LDPC terms) and λ defines the number of blocks in which each pair of points are together. Translating these parameters to the matrix field, v represents the number of columns, r is the number of 1's per column and $\lambda=1$ will determine the absence of length-four cycles: a value of 1 ensures that any pair of 1's will be in only one row.

The BIBD-based matrices chosen are rate 1/2 matrices, so, for rate 1/2 the construction is immediate, turning the BIBD into a matrix. There are several possibilities for BIBDs with $\lambda=1$. The most well known are affine planes and projective planes. The latter will be used in our designs because their total number of points is less than that needed in affine planes for a given maximum degree of parity matrix. Using fewer points allows us to use bigger permutation matrices, which leads to better BER performance and which also needs fewer memory elements in implementation.

The parameters that define projective planes are: $(n^2+n+1, n+1, 1)$. For example, Figure 1 shows a BIBD(7,3,1). The way the matrix rows defined by the BIBD are arranged should be noted.

In order to get rates lower than 1/2 the BIBD to be used is the same as the one used for 1/2, chosen to satisfy the optimal bit and check degree. The only difference is that once the top matrix has been defined, some columns are removed to adjust the rate.

For rates higher than 1/2 the process proposed consists in removing some rows of the original matrix. An example can be seen in section II-C.

The number of memories used to store information bits depends on the parallelism desired. If a high parallelism is desired, the number required to implement a BIBD with high rate (the highest parallelism is obtained using a memory for any column of the top matrix) can be a drawback for designers. For example, for a bit degree of 13 the system would require 157 memories.

In order to offer an alternative solution to this problem, another option is proposed. Depending on the bit degree desired a BIBD with a suitable number of elements per column is chosen. Moreover, depending on the rate to be reached, either the whole BIBD or only a part of it (this last option allows higher rates) is used. After that, the resulting BIBD is cloned and put side by side with itself. The only difference between both parts is the way their pseudo-random primitive generators are

designed. The primitive generators of the BIBD on the right are exactly the same as their equivalents on the left, but the *init_value* (explained in the next section) is changed depending on the row. This ensures that the new structure is still length-four cycle free with the bit degree desired and high rate. In sub-section II-C an example of design is explained.

Although all the proposed schemes define semi-regular eIRAs (H_1 been regular), the method proposed is flexible and irregular eIRAs can be obtained by removing permutation matrices in top level architecture in columns where degree should be reduced.

2.2 BOTTOM Level Architecture

Once all the sub-matrices have been emplaced using BIBDs, the way the permutation of the identity matrix is defined will be explained. The objective of primitive generators will be to produce the pointers to the systematic bits needed to generate a given parity bit.

Obviously, a completely random generator cannot be implemented. The proposed method is based on primitive interleavers (we will call them primitive generators) (Morelos-Zaragoza, 2002). These interleavers have the desirable property of being very simple to design, fast and with low area cost and, from the point of view of BER performance, they perform only slightly worse than a completely random one in this kind of systems. The interleaved positions are calculated through the following equation:

$$i_{k+1} = (i_k + root) \bmod N \quad (1)$$

There are three parameters that define the pseudo-random sequence, N (the maximum value, and the total number of values where the sequence is complete), the *init_value* (the initial value i_0) and the *root* (the difference between one value and the next or previous one). How these parameters will be chosen is important in the BER performance of the system.

The proposed design suggests an N which is prime. This fact allows the use of as many different roots as $N-1$ (a valid root is a root that forms complete N sequences). Moreover, N is chosen to satisfy:

$$framesize \approx N * column_number$$

Fortunately, despite this restriction, there are enough prime numbers and so widely distributed that any frame size can be approximated.

The idea of having several roots increments the randomness of the whole parity check matrix because despite using the same generator the sequences obtained are different. On the other hand, following recommendations in (Yang, 2004)(Dinoi, 2005)(Tian, 2004), in order to reduce low-weight codewords and near codewords, for lower degree variables, which are mainly responsible of error floors, roots are selected so that ones of columns and pairs of columns are widely separated. This is done by using an algorithm that performs two steps:

1.- For a given bit degree distribution, the initial BIBD has a column degree which is equal to maximum bit degree. Lower degrees are obtained by removing column elements. This elimination is performed in columns whose elements are less distributed.

2.-Once the optimal degree is obtained, the algorithm chooses the optimal roots for the lowest weight columns. These optimal roots will be chosen to maximize the average distance between elements within any two columns as well as in any column itself.

Moreover, the chosen root is also used as *init_value* in order to begin at a different number for any sub-matrix. Roots selected with the mentioned methodology will be called distributed roots in the next results section.

Using the above recommendations the BER performance of the system is close to that obtained using a completely random generator.

2.3 Example of Code Construction

To clarify the proposed method a simple example will be discussed. We will describe the design process for three different rates.

Rate 1/2: Suppose a rate 0.5 eIRA with column degree 3. With only these requirements the design is as easy as choosing a BIBD (7,3,1). Figure 2 displays the TOP matrix, which is clearly length-four cycle free.

P	P	P				
P			P	P		
P					P	P
	P		P	P		
	P				P	
		P	P			P
		P		P	P	

Figure 2: BIBD(7,3,1) for H₁ of rate 0.5 eIRA.

P	P	P			
P			F	P	
	P		F		P
		P		P	P

Figure 3: H₁ of rate 0.6 eIRA.

The next step would be to define the primitive generator parameters for the required generators, in the way that has been defined in II-B.

Rate 0.6: to obtain a slightly higher rate, the same BIBD shown in Figure 2 can be used. This time only four of its rows are to be used. If rows 1,2,4 and 7 are used and the last column is removed, a 0.6 rate eIRA code is obtained (with a degree of 2 for any bit in H₁). The matrix mentioned can be seen in Figure 3. Of course, depending on the degree needed the user can define which rows are used and can even add or remove columns or P's (while the length-four cycle free structure is maintained). The initial BIBD defines the maximum degree available, so if a bigger degree is needed, another BIBD, with bigger *r*, has to be chosen.

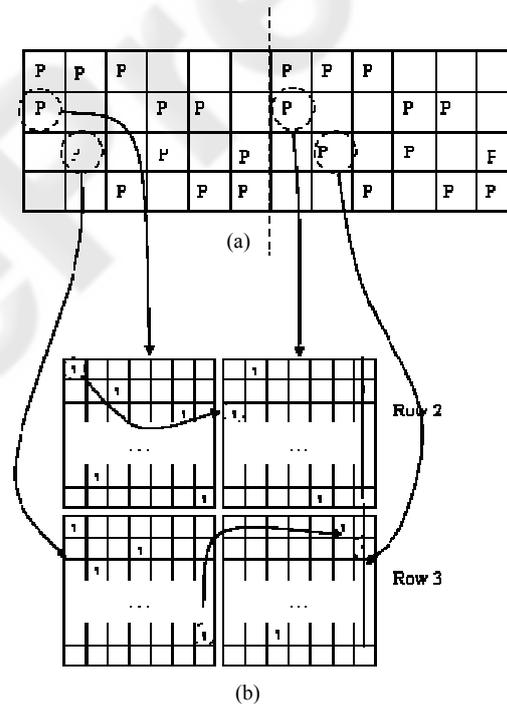


Figure 4: (a) H₁ of rate 0.75 eIRA (b) Bottom level architecture detail for 2P matrices on the left TOP level matrix and their corresponding matrices on the right.

Rate 0.75: the last example will achieve a 0.75 rate eIRA maintaining the degree of 2 for any bit. At

this point there are two possibilities as was explained previously: on the one hand, a suitable BIBD can be used, removing rows till the rate desired is obtained. On the other hand, the architecture explained in II-A can be used, i.e., choosing a smaller BIBD and cloning it to form a TOP architecture with higher rate.

For this example the second alternative is applied, using the original BIBD(7,3,1) matrix. To do so, the matrix in Figure 3 will be cloned. The primitive generator used on the left is designed in the same way as in the case of low rate. The primitive generator used on the right of the dotted line in Figure 4-a has as its *init value* the row in which that generator is placed. With this simple idea a length-four cycle free parity check matrix is obtained in spite of having length-four cycles in the TOP architecture.

The top level architecture defined in this way is displayed in Figure 4-a and a bottom level detail for two generic square P matrices is shown in Figure 4-b.

3 BER PERFORMANCE

The results obtained in the original eIRA paper (Yang, 2004) will be taken as the main point of reference. The same two rates and frame sizes reported in this paper have been tested with our methodology. Moreover, the same bit and check degrees will be used too, because they have been demonstrated to be optimal using Gaussian approximation (Richardson, 2001). The Mansour (Mansour, 2003) results will also be compared but not forgetting that it is not an eIRA approach.

3.1 Rate 0.5

For the rate 0.5 example in (Yang, 2004), the frame size used (4018, 2009) is approximately the same as the one reported there (4000, 2000). The reason for the slight frame size difference is the use of a prime N and a BIBD (49, 7, 1). In this particular case, N was set to 41 as this is the prime value that provides the frame length closest to the desired one: $41 \times 49 = 2009$.

We began using the same check and bit degrees as the original eIRA because they have been demonstrated to be optimal using differential evolution. The proposed bit degree for H_1 matrix was 58% of information bits with degree 3 and 42% with degree 7. On the other hand roots of primitive generators were randomly elected. Results can be

seen in Figure 5 labeled as *eIRA BIBD (58%w3, 42%w7, rr)* where *rr* means random roots. Performance is clearly improved selecting roots following the previously mentioned criteria, based on separating widely the ones on columns and pair of columns. The use of these distributed roots and its BER performance is labeled as *eIRA BIBD (58%w3, 0%w4, 42%w7, dr)*, where *dr* means distributed roots, in Figure 5.

The next step to improve the BER performance in the error floor zone was to increase top level columns to degree-4. A top level column is a column of the top level matrix, which contains 41 information bits in this particular case. In order to low the error floor a method based on increasing the top level columns that are involved in most low weight codewords and near codewords is proposed. Basically the method consist in studying the quantity of errors in which each top level column is involved in and increase the degree of those with most errors (Pérez, 2005).

Increasing 3 top level columns (3*41 information bits), which constitutes 6% of the total number of columns, the BER performance in the error floor zone (SNR=1.6dB) is improved from $2 \cdot 10^{-5}$ to $9 \cdot 10^{-6}$. Finally, by increasing 6 top level columns (12% of the weight 3 columns) the BER performance goes below $3 \cdot 10^{-6}$ as can be seen in Figure 5.

Final results are labeled as *BIBD (46%w3, 12%w4, 42%w7, dr)* in Figure 5, indicating the percentage of columns increased to degree-4. This final result can also be seen in Figure 6, compared to the original eIRA results presented in (Yang, 2004). As can be observed, the proposed method is really close to the original eIRA in terms of BER performance, but eliminating the random topology of the parity check matrix with the implementation benefits this feature implies.

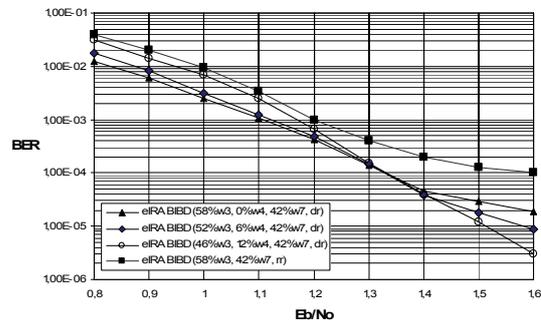


Figure 5: Influence of the percentage of weight 4 columns in BER performance.

3.2 Rate 0.8

For this scheme, the frame size was a little bigger (4495, 3534) than the one used in (Yang, 2004) (4161, 3430). Moreover, the rate is a little smaller (0.786 instead of 0.82). These two small differences can explain the performance of our system being almost 0.2dB better than the original one (using the same check and node degrees) as can be seen in Figure 7. Anyway, the difference is minimal and the performance is as close to the (Yang, 2004) design as expected.

Finally we present a comparison between Mansour (Mansour, 2003) design, a random regular design also reported in (Mansour, 2003) and our proposal for rate 0.5 and frame size (1008). In (Mansour, 2003) the degree used by Mansour was not specified.

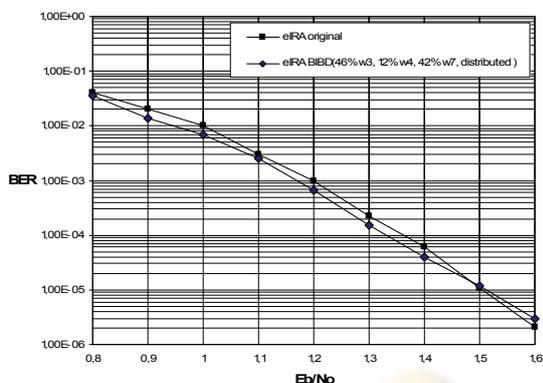


Figure 6: Performance comparison of original and proposed $n \approx 4000$ rate-0.5 eIRAs.

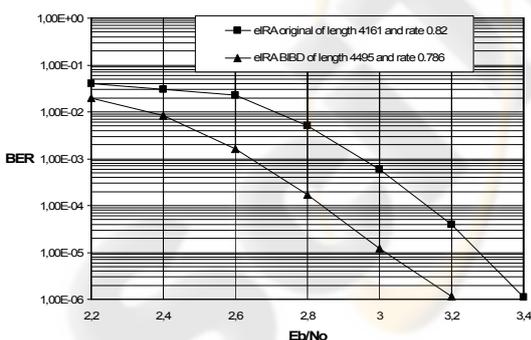


Figure 7: Performance comparison of original and proposed $n \approx 4000$ rate-0.8 eIRAs.

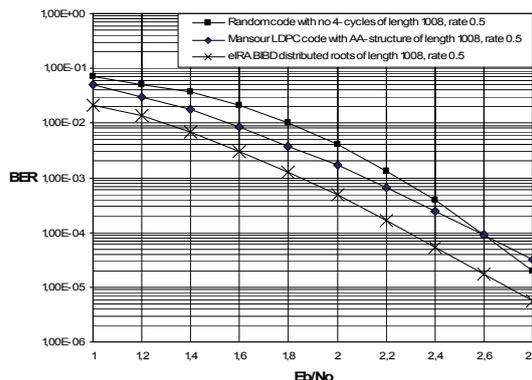


Figure 8: Performance comparison of various $n \approx 1000$ rate-0.5 regular LDPC codes.

In this paper the optimal degree calculated via the density function for this rate and size has been used.

Comparison can be seen in Figure 8. Our design outperforms both the Mansour and random designs by approximately 0.2 dB.

4 CONCLUSIONS

In this paper a new architecture for designing eIRAs is proposed. A design based on hierarchical matrices is proposed, combining the deterministic structure of BIBD $(x, y, 1)$ designs, avoiding length-four cycles, with the good BER properties of pseudo-random constructions in order to create a hardware-aware design which allows high parallelism with BER performance close to the eIRAs theoretical performance. Moreover, with the method defined in this paper we are able to improve BER performance and error floor by distributing the elements of H matrix and by selectively increasing bit degrees.

REFERENCES

Gallager, R. G., 1962. Low-density parity-check codes. In *IRE Transactions in Information Theory*, vol. IT-8, pp.21-28.

MacKay, D. J. C., 1999. Good error-correcting codes based on very sparse matrices. In *IEEE Transactions on Information Theory*, vol. 45, pp. 339-431.

DVB-S2 Standard, 2004: *EN 302 3027 v1.1.1*.

MacKay, D. J. C. and Neal R. M., 1997. Near Shannon limit performance of low density parity check codes. In *Electronic Letters*, vol. 33, no. 6, pp.457-458.

Kou, Y., Lin, S. and Fossorier M. P. C., 2001. Low-density parity-check codes based on finite geometries:

- A rediscovery and new results. In *IEEE Transactions on Information Theory*, vol. 47, pp. 2711-2736.
- Yang, M., Ryan, W. E. and Li, Y., 2004. Design of Efficiently-Encodable Moderate-Length High-Rate Irregular LDPC Codes. In *IEEE Transactions on communications.*, vol. 52, pp.564-571.
- Mansour, M. and Shanbhag, N., 2003. High-throughput LDPC decoders. In *IEEE Transactions on VLSI Systems.*, pp.976-996.
- Mansour, M., 2004. High-Performance Decoders for Regular and Irregular Repeat-Accumulate Codes. In *Proceedings IEEE GlobeCOM*, Dallas, USA.
- Liao, E., Yeo, E. and Nikolic, B., 2004. Low-density parity-check code constructions for hardware implementation. In *Proceedings IEEE ICC*, Paris, France.
- Rosenthal J. and Vontobel P.O., 2000. Constructions of LDPC Codes Using Ramanujan graphs and Ideas for Margulis. In *Proceedings of the 38-th Allerton Conference on Communications, Control and Computing*, Illinois, USA.
- Mansour, M. and Shanbhag, N., 2002. Low Power VLSI Decoder Architectures for LDPC Codes. In *Proceedings ISLPED*.
- MacKay, D. J. C and Postol, M. S., 2003. Weaknesses of Margulis and Ramanujan-Margulis low-density parity-check codes. In *Electronic Notes in Theoretical Computer Science*.
- Zhang, T. and Parhi, K. K., 2004. Joint (3,k)-Regular LDPC Code and Decoder/Encoder Design. In *IEEE Transactions on Signal Processing*, vol. 52, no. 4, pp. 1065-1079.
- Ammar, B., Honary, B., Kou, Y., Xu, J., and Lin, S., 2004. Construction of Low Density Parity Check Codes based on balance incomplete block designs. In *IEEE Transactions on Information Theory*.
- Ammar, B., Honary, B., Kou, Y., and Lin, S., 2002. Construction of Low Density Parity Check Codes: A Combinatoric Design Approach. In *Proceedings IEEE International Symposium on Information Theory*, Lausanne, Switzerland.
- Dinai, L., Sottile, F., Benedetto, S., 2005. Design of Variable-rate Irregular LDPC Codes with Low Error Floor. In *Proceedings IEEE ICC*, Seoul, South Korea.
- Morelos-Zaragoza, R. 2002. *The Art of Error Correcting Coding*. The John Wiley & Sons.
- Prabhakar, A., Narayanan, K. 2002. Pseudorandom construction of low-density parity-check codes using linear congruential sequences. In *IEEE Transactions on Communications*, vol. 50, no. 9, pp. 1389-1396.
- Tian, T., Jones, C., Villasenor, J. D. and Wesel, R. D., 2004. Selective avoidance of cycles in irregular LDPC code construction. In *IEEE Transactions on Communications.*, vol. 52, pp. 1242-1247.
- Mao, Y. and Banihashemi, H. 2001. A heuristic search for good low-density parity-check codes at short block lengths. In *Proceedings IEEE ICC*, Helsinki, Finland.
- Anderson, I. 1990. *Combinatorial Designs: Construction Methods*. The Chischester, U.K.: Ellis Horwood.
- Richardson, T. J., Shokrollahi, M. A. and Urbarke R. L. 2001. Design of capacity-approaching irregular low-density parity check codes. In *IEEE Transactions on Information Theory*, vol. 47, pp. 619-637.
- Pérez, J.M. and Fernández, V. 2005. Hardware Aware eIRA LDPC Code Generation. In *Proceedings IEEE ISWCS*, Siena, Italy.