# European Project Space on Information and Communication Systems

Angers, France February, 2015

Christophe Angot and Olivier Camp (Eds.)

Sponsored and Organized by INSTICC Published by SCITEPRESS

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Edited by Christophe Angot and Olivier Camp

Printed in Portugal ISBN: 978-989-758-155-7 Depósito Legal: 398630/15

# Foreword

This book contains the revised and extended versions of papers describing a number of European projects that were presented at the European Project Space (EPS) event organized in Angers, February 2015, associated with the set of conferences PECCS (International Conference on Pervasive and Embedded Computing and Communication Systems), PhyCS (International Conference on Physiological Computing Systems), SENSORNETS (International Conference on Sensor Networks), ICISSP (International Conference on Information Systems Security and Privacy) and MODELSWARD (International Conference on Model-Driven Engineering and Software Development).

All these events were sponsored by the Institute for Systems and Technologies of Information, Control and Communication (INSTICC) in collaboration with several international associations and other scientific partners.

The objective of the EPS is to provide a platform and meeting point for researchers and practitioners to discuss and exchange experiences on European Research and Development projects which are funded and developed in the European research area or in collaboration with European partners, thus providing insights into cutting edge research work originating from Academia or Industry, either in Europe or elsewhere.

We aim to establish the EPS as a regular event to create opportunities for knowledge and technology sharing, and establish the basis for future collaboration networks involving current project partners and interested conference delegates.

This event included a panel discussion with representatives and experts from the European Community, namely Mr Julien Van Simaeys (European Research Project Manager, University of Angers, France), whom we would like to thank for his contribution for the understanding in relation to the Horizon 2020 framework. Additionally there was also the presence of Regis Hamelin, Yole development, France and Dajana Cassioli, University of L'Aquila, Italy.

The EPS technical program included, in addition to an opening panel, the presentation of three projects which, after the event, have been invited to publish a short report in this EPS book.

We would like to thank the project representatives that decided to take their time and effort to respond to our invitation, whose reports correspond to the seven chapters of this book.

We would like to express our thanks to the EPS project representatives, who presented their projects in Angers and took the time to write the chapters of this book, and whose quality work is the essence of the EPS event and of this publication.

> Christophe Angot Olivier Camp Editors

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# Organization

# **Panel Chairs**

Christophe Angot, Angers Technopole, France Olivier Camp, ESEO, France

# **Panel Participants**

Dajana Cassioli, University of L'Aquila, Italy Regis Hamelin, Yole development, France Julien Van Simaeys, European Research Project Manager, University of Angers, France

# **Presented Projects**

Name: Research Into Wireless Channels for Intra-vehicle Communication and Positioning Presenter: Aniruddha Chandra, Brno University of Technology, Czech Republic

Acronym: MindSee Presenter: Oswald Barral, University of Helsinki, Finland

Acronym: FIRST Presenter: Zita Kovacs, University of Debrecen, Hungary

Acronym: VISION Presenter: Dajana Cassioli, University of L'Aquila, Italy

Name: EU-backed project on water quality monitoring Presenter: Ákos Milánkovich, BME-Infokom Innovátor Nonprofit Kft., Hungary

Acronym: ILLBuster Presenter: Davide Maiorca, University of Cagliari, Italy

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# Papers

# **Intra Vehicular Wireless Channel Measurements**

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Abstract. Intra vehicular communication is an emerging field of research due to its potential applications in safety of passengers, navigational and localization aids for driver, and multimedia infotainment. This chapter describes the wireless channel sounding activities performed inside a typical passenger car under the SoMoPro and the GACR projects. Three different channel sounding architectures are developed for the purpose, namely direct pulse based sounding, frequency domain sounding and pseudo noise sequence based time domain sounding. Experiments with different placements of transmitter and receiver antennas inside the car were performed for both the ultra-wide-band and millimeter wave band. Channel transfer functions and impulse responses extracted from all these measurement campaigns are utilized to construct several deterministic and statistical channel models. The models are useful for designing intra vehicular wireless links and for devising novel vehicular localization algorithms.

#### 1 Background

In recent years there has been a growing demand for wireless personal area networks (WPANs) offering high data rates for short-range indoor communication applications. The manufacturers of vehicles, aircrafts, etc. have a great interest in replacing wired communication links by wireless one in order to save installation costs.

Intra vehicular wireless links are also pivotal for intra-vehicle sensor networks. Modern vehicles employ a large number of sensors to provide vital information such as temperature, wheel rotation speed, distances to nearby objects, etc., for the electronic control units. As the number of such sensors steadily increases, the physical wires between the electronic control unit and the sensors pose significant challenges in design because wires are expensive, wiring harness belong among the heaviest components in a vehicle and have a large impact on fuel consumption, and wires are also restrictive because there are a few locations in the vehicle where sensors cannot be deployed (steering wheel, tyres, and windshields).

Another aspect for intra vehicular wireless links to be considered is the possibility of object (device or people) localization. For example, pervasive electronic gadgets require

position specific instructions (e.g. hands-free profile for driver), the safety equipments such as smart air bags require knowledge of passenger occupancy, and personalized infotainment demands parallel location oriented multimedia streaming.

For such a wide range of intra vehicular wireless applications, the prospective candidates are, ultra-wide-band (UWB), millimeter wave (mmW), and infrared (IR) or optical band technologies. UWB technologies working in the 3.1 GHz to 10.6 GHz frequency band provides data rates up to a few Gbps for the short-range communication in WPANs. On the other hand, 60 GHz mmW band (55-65 GHz) offers low latency and high transmission capacity of up to 2 Gbps. Finally, IR communication has the advantages of utilizing unregulated and unlicensed electromagnetic spectrum, offers high quality data transmission, and is immune to electromagnetic interference.

#### 1.1 GACR Project

The GACR project [1] is scheduled for four years (2013-2016) and is funded by the Czech Science Foundation. The title of the project is, 'Research into wireless channels for intra-vehicle communication and positioning'.

The main emphasis of the project is measurement and modeling of the intra-vehicle channel for WPAN application in the UWB, mmW, and optical bands and for object localization application in the UWB, and mmW bands. In addition, the existing mmW and UWB channel models derived for WPAN and sensor networks applications will be verified and compared with models created for optical bands.

The project includes six work packages as described below:

- **WP 1:** Creation of the workplace for measurement of mmW, UWB, and optical signal propagation within the vehicle.
- **WP 2:** Measurement of the intra-vehicle signal propagation for WPAN, sensor network and positioning applications.
- **WP 3:** Analysis and modeling of mmW and UWB signal propagation for WPAN and sensor network application, verification and improvement of the existing channel models.
- **WP 4:** Analysis and modeling of mmW and UWB signal propagation within the vehicle for the localization purposes.
- **WP 5:** Analysis and modeling of the IR propagation within the vehicle for WPAN application.
- **WP 6:** Verification of usability of the mmW, UWB, and optical bands for particular application and determination of accessible parameters.

#### 1.2 SoMoPro Project

The SoMoPro project [2] is a three year (2014-2016) Marie Curie COFUND activity jointly funded by European Commission under the seventh framework program (FP7) and by the region of South Moravia, Czech Republic. The title of the project is, 'Localization using ultra wide band wireless systems: from algorithms to hardware implementation'.

Localization with current global positioning system (GPS) is restricted to the areas where there is a clear line-of-sight (LoS) to satellites. Apart from that, the resolution is very poor (within several meters) for local use. While indoor localization systems do exist (ultrasound, infrared etc.) they are expensive, hard to install and maintain, and suffer from similar resolution problems.

UWB nodes are relatively cheaper, transmit low power, and are capable of accurate localization in dense cluttered environments, owing to their inherent high delay resolution and ability to penetrate obstacles. Rather than competing GPS, UWB may augment the capabilities of GPS and operate in a compatible manner. UWB can provide detection, ranging, and motion sensing of personnel and objects through walls with centimeter precision.

The overall objective of the project is to examine UWB transmission as a possible candidate for localization purposes. In order to align the research with the GACR project activities, we have chosen the vehicular environment as our primary target application area.

#### 1.3 Organization of the Chapter

After briefly reviewing the project objectives, we present the theoretical background of the three channel sounding architectures in Section 2. Section 3 deals with the measurement setup and parameters. The measurement setups and important results for experiments in UWB (3-11 GHz), mmW (55-65 GHz) and narrowband (5.8 GHz) are documented in Section 4, Section 5, and in Section 6, respectively. This chapter concludes with a description of future project activities in Section 7.

#### 2 Channel Sounding Techniques

A basic wireless channel sounding experiment refers to exciting the channel with some known radio frequency (RF) signal and measuring the response (amplitude and phase) of the channel. As physical wireless channels can be approximated as linear filters, the impulse response of the channel completely characterizes the channel [3].

Depending on the frequency range of the sounding signal, radio channel sounding can be either narrowband or wideband. Further, the radio sounding can also be grouped under either time domain or frequency domain sounding. The impulse response, h(t), obtained in Section 2.2 and Section 2.4 characterizes the channel in the time domain, while Section 2.3 characterizes the channel in the frequency domain by finding the channel transfer function, H(f), approximated with the forward transmission coefficient,  $s_{21}$ . It is possible to convert one to another through Fourier transform [4]. As the measurements are all done with digital devices, the Fourier transforms employed are the fast Fourier transform (FFT) to get the transfer function from the CIR,  $H(f) = \mathcal{F}{h(t)}$ , and the inverse fast Fourier transform (IFFT) to obtain the CIR from *s* parameters,  $h(t) = \mathcal{F}^{-1}{H(f)}$ .

#### 2.1 Channel Parameters

The field work related to our projects [1, 2] consists of various channel sounding experiments performed inside or around a passenger car parked in an underground garage. The static condition of the vehicle as well as fixed transmitter (Tx) and receiver (Rx) antenna positions allows us to neglect Doppler shift and time variations of the channel.

*Channel impulse response*: The baseband complex channel impulse response (CIR) under such assumptions may be expressed as follows

$$h(t) = \sum_{k=0}^{N} \alpha_k \exp(j\theta_k) \delta(t - \tau_k)$$
(1)

where the received signal is a composition of *N* multi-path components (MPCs), and  $\alpha_k$ ,  $\theta_k$  and  $\tau_k$  denote the path gain, phase shift and delay of the *k*th path, respectively. Unless used for localization purposes, the measured delay for the first arriving line of sight (LoS) path is set to zero, i.e.  $\tau_0 = 0$ , so that the delays for other non line of sight (nLoS) paths, i.e.  $\tau_k$ ;  $k \neq 0$ , may be termed as excess delays.

*Power delay profile*: The power delay profile (PDP) is defined as the expectation of the average received power as a function of delay time when an impulse is transmitted. It is closely related with the CIR [5]

$$PDP(t) = \mathcal{E}\left\{|h(t)|^2\right\}$$
(2)

For comparison, the obtained PDPs are often described in the normalized form

$$PDP_n(t) = PDP(t) / \int_0^{\tau_{max}} PDP(t) dt$$
(3)

where  $\tau_{\max} = \max_{k}(\tau_k)$  denotes the maximum excess delay.

*Mean excess delay*: As the name implies, mean excess delay, denotes the weighted average delay and may be found from the first moment of the PDP

$$\bar{\tau} = \int_0^{\tau_{\max}} t \cdot PDP_n(t) \, \mathrm{d}t \tag{4}$$

*RMS delay spread*: Root mean square (RMS) delay spread is the second central moment of the PDP

$$\tau_{\rm rms} = \sqrt{\int_0^{\tau_{\rm max}} (t - \bar{\tau})^2 \cdot \text{PDP}_n(t) \, \mathrm{d}t}$$
(5)

#### 2.2 Direct Time Domain Sounding

If the channel is assumed to be linear and time invariant, the received signal, r(t) = s(t) \* h(t), may be expressed as the convolution of the transmitted signal s(t) and the channel impulse response h(t). Thus, the most straightforward way to find the channel impulse response is to send an impulse as the transmitted signal,  $s(t) = \delta(t)$ , which yields a CIR

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at the receiver according to,  $r(t) = \delta(t) * h(t) = h(t)$ . However, the generation of an ideal impulse is not possible, and in practice an impulse like waveform with narrow pulse width is transmitted. It may be mentioned here that this method is primarily suitable for UWB measurements.

The main advantage of the pulsed sounding technique is, the channel impulse response is recorded in real time. In general, a repetitive pulse train is used for such sounding which necessitates fast acquisition at the receiver side. If a digital sampling oscilloscope (DSO) is used for this purpose, it should be able to operate at sampling rates of 20 Gs/s [6]. In our experiments we did not employ periodic pulse sounding as we did not try to observe the time variance of the channel.

There are several drawbacks associated with this method. First, the method needs special function generators. In fact, we abandoned this approach after first few set of experiments due to hardware unavailability. Second, impulsive signals are difficult to amplify due to RF amplifier nonlinearities [7]. Third, the poor dynamic range limits application of this method for larger Tx-Rx seprations.

#### 2.3 Frequency Domain Sounding

Frequency domain channel sounding is generally implemented through a vector network analyzer (VNA). The VNA uses a stepped frequency sweep to measure the channel in the frequency domain and records the forward transmission parameters. For a simple two port VNA this boils down to the recording of  $s_{21}$ , when the Tx and Rx are connected to port 1 and port 2, respectively.

The main benefits of the VNA are its large dynamic range, flexible frequency control, and smooth hardware synchronizations. Also the same setup may be used for both narrowband and wideband sounding by changing simple settings of the VNA. The requirement that Tx and Rx antennas should be within cable length and the channel to be static are also satisfied for in-car sounding experiments.

However, the VNA systems suffer from a slow measurement time [8]. One should also keep in mind that the CIR obtained with VNA through the IFFT operation is,  $h_{\text{VNA}}(t) = h(t) * h_{\text{fil}}(t)$ , where  $H_{\text{fil}}(f) = \mathcal{F}\{h_{\text{fil}}(\tau)\}$  is the transfer function of the windowing operation.

#### 2.4 PN Sequence based Time Domain Sounding

Pseudo noise (PN) sequence based time domain sounder use the following principle [4, 9]: if white noise, n(t), is fed as input to the channel, and the received signal,  $r(t) = \int h(\tau)n(t-\tau)d\tau$ , is cross correlated with a delayed version of the input, the correlator output

$$\mathcal{E}\{r(t) \cdot n^{*}(t-\tau)\} = \mathcal{E}\left\{\int h(\xi) \cdot n(t-\xi) \cdot n^{*}(t-\tau)d\xi\right\}$$
  
= 
$$\int h(\xi) \cdot R_{n}(\tau-\xi)d\xi = N_{0} \cdot h(\tau)$$
 (6)

is proportional to the impulse response of the channel. In the above set of equations,  $R_n(\tau)$  is the autocorrelation function of white noise n(t), which is equal to the single-

sided noise power spectral density,  $N_0$ . A PN sequence is a long sequence having noise like properties.

In our experiment, we have used maximal length sequences or *m*-sequences as pseudo random binary sequences (PRBSs) due to their good autocorrelation properties. At the transmitter, a correlation sounder is thus composed of a PN sequence generator and at the receiver data may be recorded with a DSO. It is possible to realize a matched filter in the DSO matched to the specific *m*-sequence that was generated and obtain the impulse response in real time. We prefered to do the processing off-line with MATLAB after collecting the received signal samples from the DSO.

The PN sequence based method overcomes the low dynamic range problem of direct time domain sounding due to the inherent processing gain achieved by the cross correlation process. Correlation processing also suppresses narrowband interference signals. The post-processing is, however, a bit complex, and an accurate synchronization between Tx and Rx is needed [6].

#### **3** Measurements Setup and Parameters

#### 3.1 Measurement Parameters

There are some important measurement parameters, as listed below, which affects measurement accuracy and speed.

Frequency Range and Bandwidth: In VNA based measurements, the entire frequency range between a start frequency  $(f_L)$  and a stop frequency  $(f_H)$  is swept. The number of discrete frequency tones generated by the VNA  $(N_{\text{VNA}})$  in the range and the bandwidth,  $BW = f_H - f_L$ , determine the frequency resolution

$$f_s = (f_H - f_L) / (N_{\text{VNA}} - 1) = \mathbf{BW} / (N_{\text{VNA}} - 1)$$
(7)

On the other hand, the frequency response of an impulse like wave spans over from DC ( $f_L = 0$ ) to a high cut-off frequency ( $f_H$ ). The bandwidth, BW =  $2/t_d$ , is determined by the duration of the pulse,  $t_d$ . For the PN sequence based setup,  $t_d$  should be replaced with  $T_c$ , the chip duration.

*Time and Distance Resolution*: For VNA, the bandwidth also determines the time resolution or the minimum time between samples in the CIR function obtained after IFFT

$$t_{\rm res} = 1/BW \tag{8}$$

The distance resolution refers to the length an electromagnetic wave can propagate in free space ( $c = 3 \times 10^8 \text{m/s}$ ) during time  $t_{\text{res}}$ 

$$d_{\rm res} = c \cdot t_{\rm res} = c/\rm BW \tag{9}$$

For direct time domain pulse sounding, the duration of the pulse  $(t_d)$  sets the time resolution, with minimum resolvable delay between MPCs being equal to pulse duration. If the pulse is narrower it is possible to resolve two close MPCs. Quite naturally, the chip duration  $(T_c)$  determines the time resolution for a PN sequence based setup.

*Maximum CIR Length*: The frequency step size  $(f_s)$  of the VNA determines the maximum observable delay spread, i.e. the maximum time delay until the MPCs are observed and the corresponding distance range [5] are

$$L_{\operatorname{CIR}(t)} = 1/f_s$$
 and  $L_{\operatorname{CIR}(d)} = c/f_s$  (10)

On the other hand, if only a single pulse is transmitted,  $L_{\text{CIR}(t)}$  is limited by the storage capacity of the receiver oscilloscope. For time domain periodic pulse sounding, the repetition rate determines the maximum unambiguous delay spread or impulse response length [9]. Thus

$$L_{\text{CIR}(t)} = N_{\text{PN}} \cdot T_c \quad \text{and} \quad L_{\text{CIR}(d)} = c \cdot N_{\text{PN}} \cdot T_c$$
(11)

are the maximum unambiguous range for PN sequence based setups. It may be noted that for *m*-sequences the length of PN sequence,  $N_{PN} = 2^k - 1$ , where k is a positive integer.

*Dynamic Range*: The dynamic range is defined by the difference between the largest and smallest amplitudes of the received multipath components

$$DR = 20\log_{10} \left[ \max\{h(t)\} / \min\{h(t)\} \right]$$
(12)

which can be directly evaluated from the CIR. The minimum amplitude of h(t) refers to the value still observed above the noise floor. Alternately, a ratio between the peak amplitude and noise floor can be used to measure DR. This definition is generally used for VNA measurements.

For a PN sequence based sounder, the DR may be found from [10]

$$DR = 10\log_{10}(N_{PN}) = 10\log_{10}(2^k - 1)$$
(13)

This is also known as the processing gain (PG).

In frequency domain sounding, there are certain other parameters of interest such as intermediate frequency filter bandwidth (BW<sub>IF</sub>) and output transmit power of the VNA ( $P_{VNA}$ ). If  $P_{VNA}$  is set to 0 dBm, the *s* parameters correspond to the received power measured in dBm. Further,  $P_{VNA}$  and the noise floor together define the dynamic range. On the other hand, by reducing BW<sub>IF</sub> the measurement accuracy can be improved. The cost paid is the increase in the measurement time.

#### 3.2 Test Vehicle and Parking Lot Environment

The vehicle under study is a right-hand drive, regular four-door sedan Skoda Octavia III (model 1.8 TSI Combi) with dimensions 4.659m (length)  $\times$  1.814m (width)  $\times$  1.462m (height), which was parked six storeys beneath ground level in the multi-floored underground garage of the Faculty of Electrical Engineering (FEKT), Brno University of Technology (VUT). Reinforced concrete walls and floors of the garage provided us with an environment that was free from any narrowband interference (e.g. WiFi, Cellular). Also, there were no other cars parked in close vicinity [11].



Fig. 1. Underground garage where the measurements were conducted (left) and test vehicle (right).

#### 3.3 Tx/Rx Antennas and their Placement

Vertically polarized monopole conical antennas were used for UWB sounding experiments. As suggested by Fig. 2, the radiation pattern exhibits omnidirectional characteristics over H-plane which is invariant in the frequency band of interest. Since the radiation pattern of the conical monopole antenna [12] is very close to the omnidirectional radiation pattern, we were able to capture a maximal number of multipath components (reflected waves). Due to a variable gain in the lower half E-plane radiation pattern (elevation angle from 90° to -90°), the antennas were placed in the car compartment so that the upper half E-plane radiation pattern, which is almost constant, was used. It means that when the antenna was placed at the cabin ceiling, it was set as bottom up. However, the reflected waves arriving from Tx antenna or incident on Rx antenna at lower elevation angles might be affected by the non-ideal radiation pattern of the antennas.

For the mmW band, we used a pair of open rectangular waveguide antennas (WR 15) for transmission and reception. As seen from Fig. 2, the radiation patterns are non-uniform. Currently, some work on developing slot antennas for mmW band are in



Fig. 2. Antenna used for UWB (top) and mmW (bottom) measurements with corresponding radiation patterns.

progress which would result in more uniform patterns [13]. For both UWB and mmW measurements, in order to avoid a degradation of the measured phase accuracy due to movements of the Rx antenna, phase-stable coaxial cables were used and included in the calibration process.

During the measurements, the Rx antenna had been placed on the driver's seat and on all other seats beside and behind the driver to imitate a hand-held mobile wireless device that belongs to either the driver or to a passenger. A plastic photographic tripod (JOBY GorillaPod) was used to maintain proper height (hand to lap separation) of the Rx antenna. The tripod was useful to keep the inverted cone base of the antenna in horizontal position for the UWB band and it helped in mounting the the Rx antenna in a face-to-face orientation for mmW band.

Determination of the Tx antenna positions were governed by two parallel objectives. First, the positions should resemble possible installation site for future in-vehicle wireless systems. For example, in modern cars the back roof may serve as a wireless docking station because this place usually contains some wiring and antenna for audio system or GPS. The second goal is to realize both LoS and nLoS scenarios.

#### 4 Measurements in the Ultra Wide Band

The UWB measurements formed the majority of our activities as the the experiments overlaps with the activity plan for both the projects, GACR [1] and SoMoPro [2]. We have developed 3 different channel sounding set-ups using the available off-the-shelf hardware. The first set-up reflects a basic time-domain approach and involves pulse sounding. To improve the dynamic range, another time-domain set-up based on PN sequence was later developed. The third set-up pertains to frequency domain sounding and was realized using a VNA.

#### 4.1 Direct Time Domain Measurements

The UWB time domain channel sounding measurements were performed inside the passenger compartment of the car in static condition. A Gaussian sine pulse was generated through the Tektronix AWG70002A waveform generator and was amplified through a high power amplifier (HPA) before feeding the signal to a wideband conical monopole antenna. The probing pulse used for intra vehicular UWB channel sounding has the form

$$s(t) = \sqrt{2\sqrt{2}/\left(t_d\sqrt{\pi}\right)} \exp\left[-\left(t/t_d\right)^2\right] \cos(2\pi f_c t + \phi)$$
(14)

having unit energy, initial phase  $\phi = 0.6\pi$ , and an effective pulse duration of  $2t_d = 0.276$  ns on either side. The carrier frequency,  $f_c = 6.5$ GHz, was set at the middle of the FCC approved band (3GHz to 10GHz).

At the receiver side an identical conical monopole antenna is placed which receives the signal. The signal is then amplified through a low noise amplifier (LNA) and viewed/ stored in a digital sampling oscilloscope Tektronix DPO72004C. For the HPA, we used Wenteq broadband power amplifier ABP1200-01-1825 which provided a gain of around 19 dB, whereas for the LNA, a Wenteq ABL1200-08-3220 was used that had a small



Fig. 3. UWB time domain measurement setup [14].

signal gain of 32 dB and a noise figure of 2 dB. Fig. 3 depicts the interconnections of the apparatus.



Fig. 4. Antenna placement in the car [14], RED: Tx antennas, BLUE: Rx antennas.

As shown in Fig. 4, a total of 52 different Tx-Rx antenna positions, with separations ranging from 0.56m to 1.9m, were tested with different degrees of passenger occupancy. Some measurements were repeated to investigate temporal variation, which were found to be negligible. It may be noted that although the car can accommodate four persons, we could vary the passenger count (including the driver) only up to three, as one of the places was always occupied by the receiver antenna and its attachments.

The received signal, r(t), for a particular measurement can be represented as

$$r(t) = h_{Rx,Ant} * h(t) * h_{Tx,Ant} * s(t) = s_{ref}(t) * h(t)$$
(15)

where  $h_{Tx,Ant}$  and  $h_{Rx,Ant}$  are the impulse responses of the Tx and Rx antennas, and  $s_{ref}(t) = h_{Rx,Ant} * h_{Tx,Ant} * s(t)$  is the reference input *template* that was obtained by measuring the response of the input s(t) in an anechoic chamber free from reflectors/ diffrac-





Fig. 5. Extracted CIR through modified CLEAN [14].

Next, CIRs were obtained by deconvolving the received signals with the input template using the modified CLEAN algorithm. The modified CLEAN algorithm [15] was faster and more accurate than the basic CLEAN algorithm [16] as shown in [14] through a detailed statistical performance comparison over a standard IEEE 802.15.3 channel simulation testbed. A typical impulse response is shown in Fig. 5 with the Tx antenna set at the left side of the windscreen near the roof (2L in Fig. 4) and the Rx antenna is placed on a tripod on the rear passenger seat on right (4R in Fig. 4) position. In general, it was found that nLoS conditions yielded more MPCs in the CIR compared to the cases when a direct LoS path exists between Tx and Rx antenna. This was due to presence of multiple reflected and diffracted paths inside the passenger compartment. The dynamic range can be increased by decreasing the threshold of the deconvolution algorithm. However, there is a possibility that fictitious entries would appear in the CIR profile due to noise, if the threshold is set too low.

Next, the CIR profiles obtained after the postprocessing via CLEAN were utilized to extract the RMS delay spread. After analyzing RMS delay spread values for different Tx-Rx distances, it was found that there exists only a weak correlation. On the other hand, it decreased consistently with higher passenger occupancy across all different Tx-Rx settings. For example, when the TX and Rx antennas were set to positions 4R and 1L positions (refer to Fig. 4),  $\tau_{rms}$  values were 6.8880 ns, 6.3442 ns, 5.6712 ns and 4.9847 ns with no passenger, with driver (D), with driver and front passenger (D and FP) and with driver, front passenger, and the rear passenger on left (D, FP, and RPL), respectively. The reduction in delay spread can be accounted for the obstruction and absorption of several MPCs by human body.

#### 4.2 Frequency Domain Measurements

The frequency domain measurements were realized with a 4 port vector network analyzer Agilent Technologies E5071Ca inside the passenger compartment of the car. Three ports were connected to three transmitting antennas and the fourth port was connected to a receiving antenna. The multiple input single output (MISO) channel sounding setup is shown in Fig. 6, and the measurement parameters are listed in Table 1. The scattering parameters, i.e.  $s_{41}$ ,  $s_{42}$  and  $s_{43}$ , were recorded which serve as the frequency domain channel transfer functions.



Fig. 6. UWB MISO measurement setup [17].

Table 1. VNA parameters for UWB measurement.

Parameter	Description	Value
$f_L$	Start frequency	3 GHz
fн	Stop frequency	11 GHz
BW	Bandwidth	8 GHz
$N_{\rm VNA}$	Number of points	801
$f_s$	Frequency step size	10 MHz
t <sub>res</sub>	Time resolution	0.125 ns
$d_{\rm res}$	Distance resolution	3.75 cm
$L_{\text{CIR}(t)}$	Maximum CIR length (time)	100 ns
$L_{\operatorname{CIR}(d)}$	Maximum CIR length (distance)	30 m
BWIF	IF filter bandwidth	100 Hz
$P_{\rm VNA}$	Transmit power	5 dBm
$H_{\rm fil}(f)$	Windowing for IFFT	Blackmann

The Rx antenna is placed at various locations inside the car compartment, on all seats and in the boot, and the Tx) antennas are placed on the left and right side of the dash-board, top corners of the windshield and at the rear part of the ceiling. There were altogether 90 possible Tx-Rx combinations (for details of antenna placement, please refer to [17]). The channel measurements are carried out for both, LoS and nLoS scenarios. nLoS is caused by the backrest of the seats, the dash-board, and/or persons sitting inside the vehicle.

When a statistical description of the received amplitude is attempted, it was found that they obey a generalized extreme value (GEV) process of type I

$$f(x) = (1/\sigma) \exp[-z - \exp(-z)]$$
;  $z = (x - \mu)/\sigma$  (16)

with  $\mu$  being the location parameter and  $\sigma$  the distribution scale parameter. Fig. 7 shows PDF of the received signal magnitudes in dBm fitted with the GEV statistics.



Fig. 7. Received signal amplitude statistics [17].



Fig. 8. Localization of the Rx antenna using TOA [17].

The experiment was repeated for a part of the whole UWB bandwidth, from 3.168 GHz to 4.752 GHz, with a bandwidth of 1.58 GHz as specified in the band 1 of the ECMA 368 standard [18], with 159 frequency points maintaining the same frequency step size ( $f_s$ ). The goal was to test the feasibility of ranging and localization in the whole UWB and its part. The receiver antenna position was estimated through the time of arrival (TOA) technique, i.e. by finding rough distance of the Rx antenna from three Tx antennas and performing subsequent ranging in two dimension (2D).

The whole process involved four steps, namely calculation of the CIR, detection of the first incident ray, calculation of the Tx-Rx distances for all 3 Tx antennas, and finally performing the Rx antenna localization. The first ray arrival was detected through a peak search algorithm to find the strongest MPC above the noise floor. However, the drawback was, if the first or direct path is immersed in the noise floor, the algorithm fails. this situation is encountered in mostly nLoS cases where the direct path is highly attenuated. Once the first ray is detected, the corresponding distance is calculated by multiplying the delay ( $\tau_0$ ) with the traveling speed of electromagnetic wave (*c*). Finally, the RX antenna localization is achieved through the trilateration technique [18]. Using the three calculated distances this technique allows 2D localization. Fig. 8 displays the result of localization in both the bands. The Rx antenna was placed on front passenger seat. It was important to identify the seat as in most in-vehicle localization applications (e.g. smart air bag, personalized communication profile etc.) we need to know the seat of the passenger holding the electronic communication device.

The error in ranging was less when the whole UWB was utilized rather than the first band. Also the ranging errors increased considerably in presence of passengers. The possible sources of error are, difference between calibration plane and phase center of antenna, inaccurate reference measurement, inaccuracy in estimating the first ray detection due to sampling, incorrect MPC detection and variation of wave propagation velocity in nLoS conditions [17].

Next, in another set of experiments, we used a similar  $3 \times 1$  MISO setup and shifted the receiving antenna over a  $10 \times 10$  spatial grid using a polystyrene rack having a 3 cm grid distance [19]. The idea was to study the spatial channel stationarity evaluated via Pearson correlation coefficients between absolute values of measured CIRs. The device configuration and the measurement parameters listed in Table 1 were unaltered except the frequency step size which was increased to 100 MHz for a faster measurement cycle. Further, the windowing for IFFT was rectangular in this set of experiments.



Fig. 9. Map of the correlation coefficient and their histograms [19].

Results of the spatial stationarity study are plotted in Fig. 9. The plots reflects the square geometry of our polystyrene rack. For Tx 1 ( $\alpha = 1$ ), the map of correlation coefficient ( $\rho$ ) shows that the impulse response in time domain is spatially stationary with a mean value of  $\rho_1 \approx 0.55$ . This behavior is typical also for Tx 2 ( $\alpha = 2$ ) and Tx 3 ( $\alpha = 3$ ). The histograms of  $\rho_1$ ,  $\rho_2$  and  $\rho_3$  are also plotted in Fig. 9.

The complex channel transfer functions (H(f)) obtained through the various set of measurements were converted to the respective envelope delay profiles or EDPs (|h(t)|) through IFFT operation. It was observed that the EDP is composed of large-scale variations (LSV),  $\wp(t)$ , and small-scale variations (SSV),  $\xi(t)$ . Further, the LSV part can be characterized by two different exponentially decaying functions, applicable to two different range of delay spreads

$$\wp(t) = \begin{cases} \wp^1(t) & ; 0 \ge t < \tau_b \\ \wp^2(t) & ; \tau_b \ge t < \tau_{\max} \end{cases}$$
(17)



Fig. 10. Average EDP and the LSV parts [19].

where

$$\wp^1(t) = A + B\exp(-C \cdot t), \quad \wp^2(t) = D\exp(-E \cdot t), \tag{18}$$

and  $\tau_b$  refers to the breakpoint in the delay spread domain. For our case it was about 4 ns. Fig. 10 shows the two different parts of the LSV of the PDP for the link between Tx 1 and Rx.

For the SSV, there exists no deterministic model as it is random. Exploiting the maximum likelihood estimation (MLE), we parametrized the superimposed SSV,  $\xi(t) = \text{EDP}(t) - \wp(t)$ , using the GEV distribution. The compbined model of EDP was validated through a two-sample Kolmogorov-Smirnov (K-S) test.

#### 4.3 PN Sequence based Time Domain Measurements

Compared to the direct time domain setup, a PN sequence based time domain sounder is simple to realize and replicate, provides correlation gain and does not sacrifice dynamic range much. It is also generally is cheaper and faster than the VNA based frequency domain setup.

The time domain channel sounder utilizing pseudo random binary sequences (PRBS)



Fig. 11. PRBS based measurement setup [20].

for UWB was built from several off-the-shelf laboratory instruments as depicted in Fig. 11. The PRBS is generated via Anritsu signal quality analyzer MP1800A at a rate of  $f_{\rm PN} = 12.5$  Gbits/s with maximum RF output power of 13 dBm. At full rate the output

chip duration is  $T_c = 1/f_{\rm PN} = 80$  ps. Our *m* sequence used a *k* value of 11, which gives a processing gain, PG =  $10 \cdot \log_{10}(2^{31} - 1) \approx 31$  dB. There is no point in using a higher *k* value, i.e. lengthening the *m* sequence as the PDP will contain the same information with extended noise floor. Additional processing gain as well as dynamic range can be obtained by averaging the repeated trail of sequences. Also the value of  $T_c$  and *k* determine the maximum length of CIR in time and in distance domains, which according to (11) are, 163.76 ns and 49.13 m, respectively.

DSO Tektronix MSO72004C is utilized as a receiver. It provides 4 channels, 16 GHz bandwidth, 50 GS/s real time acquisition rate and 31.25 ms of data storage per channel which equals to 0.625 ms at sampling rate of 50 GS/s. As the acquisition rate of the DSO is 4 times of  $f_{\rm PN}$ , the same chip is sampled 4 times. Another point to note is, for k = 11, the time period of the *m* sequence is,  $T_p = N_{\rm PN} \cdot T_c = 163.76$  ns, and only  $N_{\rm meas} = 0.625 \text{ms}/163.76 \text{ns} = 3816$  CIRs can be captured at once. Nevertheless, using advanced triggering modes, the number can be increased. It is also possible to continuously stream the real time acquired data to a personal computer (PC).

The signal quality analyzer also provides 10 MHz reference and gating or triggering signal to the oscilloscope. These are used for synchronization purposes. Compatible HPA and LNA may be included in the Tx and Rx chains, respectively, to boost the dynamic range of the system. A PC can be used to control the instruments and interchange data to provide additional features (e.g. real-time and continuous channel sounding) but is not necessary for the data acquisition.

The received data is correlated with the same transmitted PN sequence. For faster calculation, the correlation operation in the time domain is realized as a multiplication operation in the frequency domain. This action has some additional advantages too. For UWB, the band of interest is 3 GHz to 11 GHz. The spectral components outside this band is filtered out in the frequency domain.



Fig. 12. Comparison of CIR measurements using VNA and PRBS based system [20].

In Fig. 12, a comparison of measurement results utilizing VNA and the proposed system is shown. The measurements were performed in an anechoic chamber with conical monopole antennas separated by 2 m distance from each other. Received time domain signal was amplified utilizing external LNA. The horizontal axis of Fig. 12 is expressed in spatial distance instead of time lag delay. It clearly shows the first arriving multipath component at time corresponding to the distance of 2 m. There are some other multipath components that start arriving at the distance 3.64 m, which is caused by reflections from laboratory instruments, antenna holders and feeding cables. This picture clearly shows the agreement between measurements performed by the VNA and proposed system. However, the measurement carried out by PRBS system contains some spurious peaks caused by non-linear devices in the measurement chain [21].

Recently, using this setup, UWB measurement in both the time and frequency domain for comparison purpose were also performed. The antenna positions for the both domain were set identical. Currently, the results are being analyzed.

#### 5 Measurements in the mmW Band

For time domain measurement in the millimeter wave band, it was originally proposed to use a similar setup used in Section 4.1, i.e. with an arbitrary function generator and DSO, and the signal would be up converted to the mmW band before feeding it to a compatible Tx antenna (and down converted before feeding it to the DSO). An UWB to mmW FC1005V Silversima up-down converter was procured for the purpose. However, the large group delay (1ns/GHz) prohibited us using the setup. Thus the mmW band measurements performed so far are restricted to the frequency domain which are realized with a VNA.



Fig. 13. mmW measurement setup [22].

In the first set of experiments, a 4-port vector network analyzer, Rhode and Schwarz ZVA67, was used for measuring the transmission coefficient between two mmW antennas in the frequency band 55-65 GHz. Because of the low output power available at analyzer transmitter ouput, Quinstar power amplifier QPW-50662330-C1 was utilized to raise the transmitted signal level, thus only one signal path (instead of three in the UWB band) could be measured at once. The single input single output (SISO) channel sounding setup is shown in Fig. 13, and other measurement parameters are listed in Table 2.

The places for Tx antennas are chosen to imitate handheld mobile devices that belong to a person sitting in the car. Furthermore, the Rx antennas are located to have a good coverage of vehicle space. A pair of open waveguide WR15 were used as Tx and Rx antennas. The measurement setup was calibrated for zero transmission while the waveguides were connected to each other.

Table 2. VNA parameters for mmW measurement.

Parameter	Description	Value
$f_L$	Start frequency	55 GHz
fн	Stop frequency	65 GHz
BW	Bandwidth	10 GHz
N <sub>VNA</sub>	Number of points	1001
$f_s$	Frequency step size	$10 \ \mathrm{MHz}$
t <sub>res</sub>	Time resolution	100 ps
$d_{\rm res}$	Distance resolution	3 cm
$L_{\text{CIR}(t)}$	Maximum CIR length (time)	100 ns
$L_{\operatorname{CIR}(d)}$	Maximum CIR length (distance)	30 m
BWIF	IF filter bandwidth	100 Hz
$P_{\rm VNA}$	Transmit power	5 dBm
$H_{\rm fil}(f)$	Windowing for IFFT	Hanning



Fig. 14. Comparison of mmW CIR with UWB CIR [23].

A comparison of the CIRs obtained in the UWB and mmW band, when Tx antenna was at driver seat and Rx antenna is on the left side of the dashboard, is shown in Fig. 14. It can be seen that mmW CIR profile decays faster than the UWB. Moreover, the peak (caused by first/ direct path arrival) in mmW band is more distinctive compared to the UWB band peak. This is caused by frequency dependence of absorption parameters of materials that are used in modern cars [24].

Next, a comparison of ranging accuracies in the UWB and mmW band was attempted [25]. There were a series of steps involved in order to estimate the Tx-Rx distance. First, the measured complex channel transfer function was converted to the corresonding complex CIR function through IFFT. A Hann window was applied to mitigate leakage in time domain. Then, the first peak in absolute magnitude of CIR profile was detected utilizing semi-adaptive algorithm. The algorithm works with a threshold calculated from the maximum and mean value of the absolute magnitude of CIR, and it reports the first value above that threshold. A proper peak is identified in the neighborhood of that delay spread point. The distance corresponding to the detected peak is computed by multiplying the delay value and speed of light.

The ranging experiment in an empty car showed that while the distance measurement is having an average error and standard deviation of error of 6.7 cm and 8.3 cm for UWB, the values are 1.2 cm and 4.5 cm for the mmW band. For an occupied car, a switch from UWB to mmW would cause to lower the average error from 9 cm to 2 cm and the standard deviation is reduced from 10.1 cm to 4.5 cm [23]. It may be concluded that mmW band is more suitable for precise distance measurement, probably due to favorable material properties, therefore enhancing the distinctiveness of the first arrival multipath component. However, one should keep in mind that during experiments, the antenna positions were not identical for UWB and mmW because of the physical dimensions of used antennas. The maximum difference is 29 cm, but it is less in most of the cases. Also the distance was measured manually with a ruler which might introduce some error in measurement.

In another set of experiments, we used same waveguides as antennas. The device configuration and the measurement parameters listed in Table 2 were unaltered except the windowing for IFFT which was rectangular in this set of experiments. The Rx antenna has been placed at different spatial points inside the car compartment, on all seats, trunk and in front of the seats. The Tx antenna has been placed on the left and right side of the dash-board and at the rear part of the ceiling (for details of antenna placement, please refer to [22]).

Based on the data obtained from the measurement we proposed a channel model for mmW band utilizing a similar approach as in [19], i.e. decomposition of the magnitudedelay profile into the small and large scale variations. The LSV trend has been found by the Hodrick-Prescott [26] detrending filter. A Hodrick-Prescot filter works like a moving average filter and it separates the trend and cyclical components but do not cause data loss. On the other hand, we found out that the best fit to the superimposed SSV is achieved by the GEV distribution.

There had been also subsequent mmW measurements in frequency domain using ZVA67, QPW-50662330-C1, and Rotagrip precision cross table (for antenna positioning). The aim of this measurement was to study the spatial stationarity in the mmW band. The spatial grid used for antenna positioning was set to 4 mm. The same mmW measurement with the Quinstar low noise preamplifier QLW-50754530-I2 and with application of the extended calibration technique using mmW attenuator in order to get larger SNR was also performed. The data set is currently being analyzed.

#### 6 Narrowband Measurements

Due to large popularity of the Wireless Access for the Vehicular Environment (WAVE) technology in the 5.8 GHz band, we carried out a few measurements in this narrowband domain and compared our results with UWB.

For measuring and recording the transmission coefficient between transmitter and receiver, a 4-port VNA from Agilent Technologies, E5071C [27] was used. The measurement parameters are listed in Table 3. A single input multiple output (SIMO) measurement setup was realized with one Tx and three Rx antennas. The four omni directional conical monopole antennas that were used are identical. Real and imaginary parts of the transmission coefficients ( $s_{x1}$ ) were exported to MATLAB. The frequency domain data over the entire bandwidth BW = 100 MHz were partitioned into 10 MHz bins, where each bin corresponds to a sub-channel of 802.11p. All results were transformed from the frequency domain into the time domain utilizing the IFFT with a typ-

Table 3. VNA parameters for narrowband measurement.

Parameter	Description	Value
$f_L$	Start frequency	5.775 GHz
$f_H$	Stop frequency	5.875 GHz
BW	Bandwidth	100 MHz
N <sub>VNA</sub>	Number of points	801
$f_s$	Frequency step size	0.125 MHz
tres	Time resolution	10 ns
$d_{\rm res}$	Distance resolution	3 m
$L_{\text{CIR}(t)}$	Maximum CIR length (time)	8 µs
$L_{\operatorname{CIR}(d)}$	Maximum CIR length (distance)	2.4 km
BWIF	IF filter bandwidth	100 Hz
$P_{\rm VNA}$	Transmit power	0 dBm
$H_{\rm fil}(f)$	Windowing for IFFT	Rectangular

ical rectangular window. The PDP was calculated by averaging over 10 sub-channel time domain data.

A total of 15 different Tx-Rx combinations were tested with separations ranging from 0.53 m to 3.38 m. The Tx antenna was placed at three different locations inside the vehicle and at two locations outside the vehicle. In-car Tx antenna positions were set at the right rear seat, armrest in the middle of the car, driver seat, while for locations outside the car two positions at a height 1.09 meters were chosen; one in front of the car, and the other one near the right headlamp. The Rx antennas were installed on the left and right upper edges of the windshield and on the roof in the rear part of the vehicle. The positions of antennas were chosen for realizing both LoS and nLoS scenarios.



Fig. 15. Comparison of narrowband PDP with UWB PDP [28].

We compared our results for 802.11p protocol with measurements for UWB (3 GHz - 11GHz) performed using the same VNA based setup. The number of measured points were the same, however, due to the larger BW (8 GHz) and a frequency step size of  $f_s = 100$  MHz, we have a smaller time range  $t_d = 1/f_s = 10$  ns. Thus we cannot compare the PDPs in an one-to-one basis. For analyzing LSV models we use a scaled comparison

instead. Fig. 15 shows the UWB and the 802.11p normalized PDP vs normalized time samples. For UWB, it was found that the PDP constitutes one (or a few) major peaks followed by somewhat linear decreasing slope. The reader may also note the delay and lower power for the first peak of the UWB PDP. In addition, the delay and the maximum value of the peak is more for larger distances between the transmitting and receiving antennas. However, this phenomena is not observed in narrowband 5.8 GHz measurements for 802.11p.



Fig. 16. The average trend of the PDP for narrowband measurements [28].

As far as the PDP is concerned, it includes LSV and SSV, which can be designated mathematically in the following manner:  $PDP(t) = \wp(t) + \xi(t)$ , where  $\wp(t)$  denotes LSV and  $\xi(t)$  is the SSV. We expressed the LSV with a two-term exponential model

$$\wp(t) = A \exp(B \cdot t) + C \exp(D \cdot t) \quad ; 0 < t < \tau_{\max}$$
<sup>(19)</sup>

where the first term includes power from direct and major reflected rays, and the second term, with a very low slope (close to linear) reflects the power from diffused multipath components. The LSVs for the two-term exponential model for different distance between Tx and Rx antennas of 0.53 m, 0.94 m, 1.28m, 2.03 m, and 3.38 m are shown in Fig. 16. Next, the SSVs are separated from the PDP by subtracting the LSV, and are characterized utilizing logistic, GEV, and normal distributions. Two sample K-S tests validated that the logistic distribution is optimal for in-car, whereas the GEV distribution serves better for out-of-car measurements.

#### 7 Future Activities

#### 7.1 Measurements in the IR Band

The IR channel models for the intra-vehicle environment have been published in only a handful of literature [29], and active research is going on this field worldwide. The GACR project goal was to test the viability of intra vehicular IR WPAN systems. The first few testbeds for measurement were prepared in line with the UWB and mmW setups, i.e. for frequency domain characterizations a VNA based setup was proposed with optical transmitter (OTx) and optical receiver (ORx) directly coupled with VNA measuring the band between 1 to 5 GHz. In time domain, it was proposed to realize the same with a pulse generator and an oscilloscope.



Fig. 17. IR measurement setup.

The IR measurement performed in time domain revealed some shortcomings of the IR measurement setup, e.g. limited modulation depth of the Mach-Zehnder modulator andr large coupling loses between optical fibers and optical lenses resulting in very small SNR. For the new measurements the coupling loses were reduced by the modification and improvement of the mechanical parts of the OTx and ORx lenses leading to better focusing of the optical beams into the optical fibers. A planned measurement setup with network analyzer is shown in Fig. 17. The oscilloscope probe will be used as receiver. The optical signal will be genreated from a Fabry-Perot FP 1009P laser and will be modulated by the RF signal, fed from analyzer through Wenteq ABP1200-01-1825 broadband power amplifier, with a JDSU APE microwave analog intensity modulator.

Analysis and modeling of the IR channel will be performed in the same way as in the case of UWB and mmW channel modeling. For the comparison purpose the same characteristics (PDP, excess delay and RMS delay spread, etc.) and statistics (data fitting by the proper distribution) will be studied and mathematical channel models will be created.

#### 7.2 Localization

We will continue with the development of a robust algorithm for accurate estimation of the first multipath component arrival time (based on correlation receiver, matched filter receiver, receivers using various estimators such as maximum likelihood or mean square error estimator, or sparse signal reconstruction techniques). Then we are going to investigate the spatial distributions of the power path loss in the car compartment in order to analyze the accuracy of received signal strength (RSS)-based localization technique and its possible improvement using fingerprinting method. We would also like to assess possible application of the angle of arrival (AOA)-based positioning system and estimate its positioning capability. According to the topicality and desirability of the AOA-based positioning techniques we will measure the arrival angles and analyze their statistical characteristics. All the above measurements will be performed for LoS and nLoS scenarios for an empty and an occupied car and for a different deployment of reference nodes.

#### 7.3 Other Activities

In the next year, measurement in mmW band will be performed with the newly developed slot antenna array based on the substrate integrated waveguide (SIW) technology with significantly better omnidirectional radiation pattern compared to the formerly used open waveguide. Further we are going to examine effect of non-stationary environment in the car caused by running engine or loud sound produced by the built-in audio system or by the movement of passengers on the mmW channel characteristics. We also plan to extend the measurement to assess the effect and benefit of antenna arrays through the virtual array [30] approach and would perform some measurements in different brands of car and also in other vehicles such as tram or bus.

Acknowledgements. This work was supported by the SoMoPro II programme, Project No. 3SGA5720 *Localization via UWB*, co-financed by the People Programme (Marie Curie action) of the Seventh Framework Programme (FP7) of EU according to the REA Grant Agreement No. 291782 and by the South-Moravian Region. The research is further co-financed by the Czech Science Foundation, Project No. 13-38735S *Research into wireless channels for intra-vehicle communication and positioning*, and by Czech Ministry of Education in frame of National Sustainability Program under grant LO1401. For research, infrastructure of the SIX Center was used. The generous support from Skoda a.s. Mlada Boleslav are also gratefully acknowledged.

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# Wireless Sensor Network for Water Quality Monitoring

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Abstract. This book chapter introduces a European Union-backed water quality monitoring and early warning system implemented in the cross border region of Hungary and Slovakia achieved by Hungarian and Slovak partners. The measurement system consists of buoys which carry sensors and send measurement data wirelessly to a gateway on the riverside. The gateway transmits the data to a server, where it is saved and processed. Based on the collected data, alerts can be sent in case of anomalies. Utilizing this solution, an early warning system for water pollution can be deployed, which decreases water quality monitoring costs.

## 1 Introduction



Wireless Sensor Network for Water Quality monitoring (WSN-Aqua) is a joint research and development project financed by the European Union for creating a water quality monitoring sensor network and early warning system achieved by Hungarian and Slovak partners. The project started in July 2013 and is due to end in June 2015. During the two years of the project, the partners will cooperatively develop and implement all the necessary components to create the water monitoring system. The main output of the project is a test system of 8 buoys placed in the cross border region of Hungary and Slovakia on the River Ipoly.

This book chapter is organized as follows: Section 2 defines the goals of the project; Section 3 introduces the project partners along their responsibilities. Section 4 describes the considered environmental aspects and the selected location of the first demo system. Section 5 presents the details of the implementation and Section 6 discusses research results. Finally, Section 7 concludes this book chapter.

#### 2 Aims and Outcomes of the Project

The overall goal of the joint research project is to foster intensive cooperation between Hungarian and Slovak R&D institutions to increase the economical competitiveness and capacity of cross-border regions.

The project-specific goal is to develop a sensor network which could be utilized for water quality measurements and contribute to the environmental protection of rivers and lakes and be able to function as an early warning system for the authorities in a cost efficient way. Such a system would be beneficial for both countries.

The project's subject and goals are directly harmonizing with the priorities and goals of the HUSK program, especially with "Cooperation in science, research and innovation" priority (HUSK 1101/1.2.1) and the aim to increase the economical competitiveness of the cross-border region.

The main result of the joint research project is a new product capable of measuring and monitoring water quality. To implement such a system, a radio communication protocol will be developed, a buoy will be designed and a measurement test system will be deployed as a proof of concept.

The project reinforces the social and economical integration of the cross-border region of Hungary and Slovakia, more specifically it builds partnership in the scientific research community.

The outcomes of the project include a new radio communication protocol, hardware board designs, buoy design, measurement test system on Ipoly and documentation.

#### **3** Project Partners and Responsibilities

This chapter introduces the partners and their responsibilities (work packages, WPs) in the WSN-Aqua project.

#### 3.1 BME-Infokom Innovator Nonprofit Ltd.<sup>1</sup>



BME-Infokom Innovátor Nonprofit Kft. is a nonprofit organization of technology Research and Development meeting market demands. We take on various technological challenges and carry innovative ideas into execution by utilizing our long-term R&D and project experiences, as well as professional staff. Our company offers outstanding cooperation opportunities and reliable partners, such as the Budapest University of Technology and Economics.

BME-Infokom is responsible for the following work packages:

WP2 Development of an air interface for the environmental monitoring system Wireless transmission is a critical part of the monitoring system. There will be two different radio technologies applied in the system. The nodes communicate with the central server with GPRS/UMTS data transmission. The nodes use another ISM band

<sup>&</sup>lt;sup>1</sup> www.bme-infokom.hu

radio solution to provide data transmission. This work package defines the wireless interfaces and their parameters.

WP5 Development of a wireless communication protocol for the environment protecting monitoring system

This working package defines the communication protocol that makes possible to transmit the sensory data via the air interface, defined in WP2. During the design of the protocol energy efficiency was a major factor. It is important that the protocol works as efficiently as possible. This phase specifies the data packet structure, where the different types of sensor data are considered. The various message types during communication are also to be defined.

WP7 Analysis of network topologies and alternative energy sources

The work package has two goals: first, the definition of network topologies, on which the system will operate, second, to develop a module utilizing renewable energy sources, which provides long operating time for the system. The topology largely determines the protocol being developed, so it is connected to the processes defined by WP5. The various situations require very different topological solutions, in order that we have to prepare for more scenarios.

WP8 Establishment of the system monitoring system

The system is placed open terrain, therefore it is exposed to weather effects, ship traffic and possible damage of vandals. The intermittent failure of a device is una-voidable. The aim of this WP is to create a centralized remote monitoring system that collects the status information of the devices. The information should be collected with energy efficiency taken to account. The protocol has to be implemented in a way that it uses as few messages as possible, but also provides satisfactory amount of status information.

#### 3.2 Technical University of Kosice<sup>2</sup>



Technická Univerzita V Košiciach (TUKE) was founded in 1952 and is currently active in nine scientific fields as a research and educational regional center. Currently there are more than 10000 students at the university. Besides educational activities, TUKE has successfully completed many R&D projects.

TUKE is responsible for the following work packages:

WP1 Analysis and selection of sensors for water quality monitoring in distributed sensor network nodes

During this activity, we analyze and select a set of suitable sensors to be used for cost effective monitoring of water quality in distributed highly embedded network nodes. There is a wide variety of sensors and transducers at the market, which can measure temperature, pH, oxygen ratio, water flow volume, etc. Several key parameters of selected sensors will be optimized during the selection process.

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<sup>&</sup>lt;sup>2</sup> www.tuke.sk

WP3 Sensor node hardware components, architecture and embedded software development

This activity deals with the selection of suitable hardware components, composition of the sensor node hardware in terms of used embedded devices, such as particular microprocessors, particular radio frequency transceiver chips (RF chips), various sources of energy, and also with efficient energy management.

WP4 Implementation of data confidentiality and integrity algorithms and protocols

During this activity, we analyze and implement a suitable set of cryptographic algorithms and protocols that ensure confidentiality and integrity of transmitted sensors data via the air interface, defined in the second activity and to be compatible with available communication protocol data structures defined in the fifth activity. Confidentiality and integrity of protected data will be based on a suitable combination of symmetric and asymmetric algorithms implemented on an embedded hardware platform defined in the third activity.

WP6 Development of sensor data processing methods for the comprehensive evaluation of the water quality state

The application of the sensor network to be developed within the project is to provide corresponding specialists and authorities with the relevant information on the water quality state within the monitored water flow. The activity is intent on the design of the final content and form of reporting the water quality state based on processing of data provided by the particular nodes of the sensor network.

Both partners are responsible for the following work packages:

WP9 Development and evaluation of the environmental monitoring test system This WP aims at integrating all the components created in the previous 8 WPs and also tests the whole system in operation in real-word conditions.

WP10 Dissemination of the project results

The project results can be summarized as a contribution to environmental protection (social contribution), the sensor network applied for the water quality monitoring (technical solution), the novel scientific and engineering methods applied for sensor network design and development (research results) and by the excellent example of Hungarian-Slovak cross-border cooperation. Each of these different kinds of the project results will be disseminated by the methods, such as popularization presentations, seminars, papers in journals and conference proceedings, the developed sensor network demonstration and education process.

#### **4** Environmental Aspects

Each unit of the environmental monitoring system implemented within the framework of the project will include sensors capable of measuring 4 parameters, enabling realtime processing and visual display of these data. Thanks to the communication method of the new sensor system, users (authorities, water experts) will be able to locate contaminations, physical and chemical changes or other events, and to monitor or model these in real-time, or to provide estimates and forecasts.

In this project, we examined the temperature, pH, conductivity and the level of dissolved oxygen (mg/l) in a given river sector. The sensors will provide us with five further parameters: redox potential, dissolved oxygen (%), pressure, salinity, total dissolved salts (kcl). Later on, replacing and extending the sensors can enable us to monitor the locally significant parameters of a chosen area.

The disadvantage of the current monitoring methods (measurement stations or analyzing water samples in a laboratory) is that experts receive the results with a delay, rendering immediate intervention in case of a sudden damage impossible. Another hindrance is that the presence of water experts is constantly required, but – due to the lack of resources – this is impossible to provide.

Using radio communication technology, the data of the wireless sensor network used in the project is sent to other units for processing and evaluation. Thanks to this, if the level of contamination reaches a limit, experts or authorities can be informed immediately. Although the results are of course not as accurate as in a laboratory, after a real-time alert, further analysis can be carried out in order to decide whether immediate intervention is required, which can play an important role in effectively mitigating damages to the environment.

The standalone mode of the sensor network and technology also enables hard-toreach areas to be monitored. This possibility is available thanks to the energy efficient and renewable energy based operation of the devices.

#### 4.1 Criteria for Choosing the Monitoring Area

Before choosing the monitoring area, we had examined a number of river sectors. As the project is a Hungarian-Slovak joint task, we tried to choose cross-border rivers (Hernád, Sajó) or rivers that are natural borders and their drainage basin and tributaries are present in both countries (Danube, Ipoly)

We wanted to choose a river which lacks shipping routes as ship traffic can damage the buoys, the obligatory illumination of buoys results in higher energy consumption – further increasing the size of the required solar panel –, and the authorization process takes longer.

After installation, the buoy system can be damaged by natural phenomena as well as human factors. We examined the number of police calls and the population density in the areas of the aforementioned rivers, because we wanted to choose the safest, most sparsely populated area.

In terms of natural phenomena, rivers can be most severely affected by floods. The hydrometeorological data of previous years enables us to predict the temporal distribution of floods.

It is also very important to choose a location that is easily accessible.

The aim of the monitoring system is to detect damages and contaminations that negatively affect the condition of the river. This is why we tried to choose an area which has potential water pollution sources nearby.

#### 4.2 Limit Values of Water Quality

The results of river quality analysis have to be compared with the current limit values of water quality. The water quality classes are as follows: I Excellent, II Good, III Acceptable, IV polluted, V severely polluted. In cooperation with the authorities, we

set the alert threshold for the system. Decree 10-2010 lays down the following limit values (listed in Table 1):

Annex II of Decree 10/2010. (VIII. 18.) of the Ministry of Rural Development<sup>3</sup> 1. LIMIT VALUES OF WATERS

1.1. WATER QUALITY LIMIT VALUES OF RIVERS

		Type of water body based on the relevant legislation
1	Physical and chemical characteristics	Small plain rivers (type 11, 12, 15, 18)
2	pH	6.5-9
3	Conductivity (µS/cm)	<1000
4	Chloride (mg/l)	<60
5	Oxygen saturation (%)	60-130
6	Dissolved oxygen (mg/l)	>6
7	BOI5 (mg/l)	<4
8	KOIcr (mg/l)	<30
9	NH4-N (mg/l)	<0.4
10	NO2-N (mg/l)	<0.06
11	NO3-N (mg/l)	<2
12	Total N (mg/l)	<3
13	PO4-P (mg/m3)	<200
14	Total P (mg/m3)	<400

Table 1. Water quality limit values of rivers.

We consulted with our Slovak partner about the water quality limit values set by Slovak authorities. As there is a slight difference between the limits of the two countries, we agreed on creating two different settings in terms of limit values. The first buoy cluster is set according to the Slovak limits, while the second cluster is in harmony with the Hungarian limits.

Based on this, the monitoring system alerts in the cases summarized in Table 2:

	SK – 1 <sup>st</sup> buoy cluster (average/moderate status)	HU – 2 <sup>nd</sup> buoy cluster
рН	<6 and 9<	<6.5 and 9<
conductivity [µS/cm]	>700	>1000
dissolved oxygen [mg/l]	<6.5	<6
T [°C]	>24	(>24)

Table 2. Alert limits for the monitoring system.

<sup>3</sup> http://net.jogtar.hu/jr/gen/hjegy\_doc.cgi?docid=A1000010.VM

#### 4.3 Parameters Assessed during the Project

Each buoy of the WSN-AQUA monitoring network will be equipped with 3 Ponsel sensors, which are capable of measuring pH/redox, conductivity, dissolved oxygen (mg/l) and temperature.

pH (pondus Hydrogenii): Chemical characteristic of aqueous solutions which shows the acidity or basicity of the solution. A pH of 7 is neutral. A pH less than 7 is acidic and a pH greater than 7 is basic. The pH is defined as the decimal logarithm of the reciprocal of the hydrogen ion activity. Hydronium is produced by protonation of water, in which giving up a proton results in hydroxide (OH–) and hydronium (H3O+) ions.

$$pH = -log_{10}[H_3O^+] = -lg[H_3O^+]$$

Redox potential: ORP – Oxidation Reduction Potential, a value measured in mV. It is the tendency of a chemical species to acquire electrons and thereby be reduced. The more positive the potential, the greater the species' affinity for electrons and tendency to be reduced.

Dissolved oxygen [mg/l] (OPTOD – Optical Dissolved Oxygen technology): It is an important parameter in assessing water quality because of its influence on the organisms living within a body of water.

Conductivity and temperature: Conductivity is a measure of the ability of water to pass an electrical current. For this, we use a Ponsel sensor to assess electrical current ( $\mu$ S/cm). As conductivity increases by 2% in case of a 1 °C temperature rise, the measurement has to be standardized to 25 °C.

Although all three sensors measure temperature, this factor is very important when measuring conductivity, therefore the conductivity sensor's temperature data will be considered as reference value.

#### 4.4 The Prioritized Area

After preliminary consideration, from the Ipoly, Sajó and Hernád river sections we decided to further examine the Ipoly area. We considered the security of the area, but the distance between possible locations was also a different factor as it greatly influences setup and maintenance costs, and shorter distances mean that the area can be reached faster in case of an alert. The description of the river basin management plan (RBMP) of Ipoly and a debate report for its review from 2014 provide us with useful information about the area.

The Ipoly meets the criteria of our project because it acts as a natural border for almost 100 km and the quality of the water flowing into it from its river basin and tributaries can indicate pollution along the river basin.

Due to river control, the Ipoly is now only 212.5 km long, its river basin is 5151 km2, out of which 1502 km2 is located in Hungary. The Slovak Ore Mountain, the Selmec Mountains, the Karancs, the Cserhát and the Börzsöny are all located in its river basin.

The valley of the Ipoly is characterized by basins separated by natural formations that hinder erosion. In several areas, it has a gentle gradient with tributaries of steep gradient. The river bed consists of impermeable rocks, but it is not an abundant river

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due to little rainfall and small river basin. Its fluctuation is however considerable, any significant amount of rainfall triggers flood waves that last no longer than few hours or 1-2 days maximum. The affected areas are mainly protected by embankments, dams and pumping stations.

#### 4.5 Protected Areas<sup>4</sup>

According to the EU Water Framework Directive, every area receives protected status where any legislation identifies surface waters and groundwater or water dependent species to be protected.

In the Ipoly subunit of the river basin management plan, we can come across the following categories of protected areas:

Protection of drinking water, nutrient or nitrate sensitive areas, natural bathing sites, areas protected due to their natural values.

#### 4.6 Flood Risk

The fluctuation of rivers is another important factor when installing a buoy system. The fluctuation of the Ipoly is fairly wild and there are two flooding periods during the year, one at spring snowmelt and another in summer due to heavy rainfall. The water level is low in August and September. Floods are also possible in winter due to fast snowmelt. According to observation data collected at Nógrádszakál, the lowest water level was 33 cm, while the highest was 362 cm.

The Middle-Danube Valley Water Directorate prepares monthly hydrometeorological reports of its territories. The reports provide an analysis of the meteorological situation of the given month and the temperature changes as well as the connection between these two factors and their effect on river fluctuation. These reports contain water level and discharge parameters of the Ipoly, the Zagyva and the Danube. The reports also provide information regarding groundwater level and the water management situation. The latter includes data about the condition and water level of reservoirs operated by the Directorate. There is also information regarding the level of flood preparedness in certain areas as well as water quality data (dissolved oxygen, ammonia, nitrite, nitrate, minerals and water quality classification).<sup>5</sup>

We examined the river fluctuation data of hydrometeorological reports available on the website (dating back to July 2011). Our aim was to provide an overview of river fluctuation characteristics of the Ipoly, which is a required measure before installing the buoy system.

Installing the monitoring system is due to late April and early May. As earlier observations suggest, we have to be prepared for sudden flood waves, so we intend to install the buoys considering meteorological data and hydrometeorological forecasts of the Directorate. Generally, flood waves are not frequent in the onset of summer.

<sup>&</sup>lt;sup>4</sup> http://vizeink.hu/files3/1\_8\_Ipoly.pdf – available in Hungarian

<sup>&</sup>lt;sup>5</sup> http://www.kdvvizig.hu/ Middle-Danube Valley Water Directorate

Our findings suggested that the Ipoly is the most suitable of the possible areas for our purposes. In order to set the exact location of the monitoring system, further research is required, and circumstances of carrying out the installation and maintenance have to be considered as well.

#### 4.7 Choosing the Monitoring Site, Field Inspection

In order to choose the most suitable site for the monitoring system, we carried out preliminary water quality surveys during our field inspections of the areas that were deemed suitable based on our background research. In the future, the results of these surveys can be used as reference values to compare with results measured by the monitoring system.

Firstly, we inspected three fields on the lower Ipoly section on the river sections along Letkés, Ipolydamásd and Szob. A representative of Aqua-Terra Lab Ltd., the company responsible for designing and manufacturing the buoys, and a colleague of the Institute of Environmental Engineering of Pannon University also participated in the field inspection.

During the field inspection, we assessed the accessibility of the sites as well as the suitability of the river section regarding buoy installation and maintenance. We also tested the Ponsel sensors to be used during the project, and we collected water samples for laboratory measures at Pannon University. We wanted to choose a site where contamination or external impacts that influence water quality are highly possible. This is why we tried to choose confluences of tributaries, areas close to settlements or river sections near arable lands.

After inspecting the lower Ipoly section, we considered the characteristics of all three sites and arrived at the conclusion that the area between Letkés and Salka is the most suitable site to install a monitoring system. The site is easily accessible, there is hourly gauge data available both at Ipolytölgyes and Salka, which can be followed at http://www.vizugy.hu/<sup>6</sup> and the Slovak website, http://www.shmu.sk/<sup>7</sup>.

The next step was to consult the experts of the Middle-Danube Valley Directorate. They expressed their support of the monitoring system. Since the project field is located on a river that acts as natural border between two countries, we had to contact the competent Slovak water management authority. The experts of the Directorate also helped us in this case, as they established close cooperation and strong relationship with the Slovak colleagues who are competent in this area.

After prior consultation, we inspected the Letkés section of the river and we discussed our plans regarding the buoy system. The meeting was attended by two experts of the Middle-Danube Valley Directorate, two representatives of the competent Slovak water management directorate, the leader of the river warden service of the section, two river wardens, an expert of Aqua-Terra Lab and of the Institute of Environmental Engineering of Pannon University as well as the engineers of BME-Infokom.

<sup>6 //</sup>www.vizugy.hu/

<sup>7</sup> http://www.shmu.sk/

During the meeting, we presented the aims of the project to the Slovak experts, and they also expressed their support. The Slovak and Hungarian directorates issued written permit to install the monitoring system.

During the field inspection, we also tested whether sufficient radio communication is possible in the area. The data collected from the two gateways and the measuring results show that decent data transmission is possible in the area. This means that the transfer – to gateways – of data collected by the buoy sensors will be perfectly feasible.

The monitoring points are to be installed in two separate chains because, according to the information of the Directorate, the island under the bridge at Letkés is going to be demolished next spring. Since we would like to avoid the river section affected by the earthworks, we intend to install two set of buoys (in an upper and a lower section).

#### 4.8 Devices used for On-the-Spot Measurement

The following sensors were used in the surveyed areas to assess temperature, pH, redox potential, conductivity, dissolved oxygen, oxygen saturation and turbidity:

NOETEK-PONSEL pH/redox measuring instrument: pH, redox potential

NOETEK-PONSEL OPTOD optical dissolved oxygen measuring instrument: dissolved oxygen, oxygen saturation

NOETEK-PONSEL C4E: temperature, conductivity

NOETEK-PONSEL digital nephelometric turbidity measuring instrument: turbidity (NTU)

#### 4.9 Standards and Regulations Followed

MSZ EN ISO 5667-1:2007 Water quality. Sampling. Part 1: Guidance on the design of sampling programmes and sampling techniques

MSZ EN ISO 5667-3:2004 Water quality. Sampling. Part 3: Guidance on the preservation and handling of water samples

MSZ EN ISO 5667-6:1995 Water quality. Sampling. Part 6: Guidance on sampling of rivers and streams (superseded standard)

MSZ 448-2:1967 Drinking water analysis. Determination of temperature, colour, turbidity and limpidity (superseded standard)

MSZ EN ISO 10523:2012 Water quality. Determination of pH (ISO 10523:2008)

MSZ 448-32:1977 Drinking water analysis. Determination of specific electrical conductivity (superseded standard)

MSZ EN 25814:1998 Water quality. Determination of dissolved oxygen. Electrochemical probe method

MSZ EN ISO 7027:2000 Water quality. Determination of turbidity

#### 4.10 Code Table of Colour, Odour and Weather of Surface Waters

Simplified summary (see **Table 3**) of categories and codes used when sampling surface waters (exception: Lake Balaton and its surrounding rivers).

Table 3. Code tables for categories.

Weather	
First digit	Precipitation
no precipitation	0
humid	1
foggy	2
rain	3
shower, thunderstorm	4
ice pellets	5
snow	6
snow shower	7
hail	8

Colour

First digit	Colour strength
colourless	0
colour intensity	1-7
Second digit	Dominant colour
colourless	0
yellow	1
green	2
blue	3
grey	4
brown	5
black	6
red	7

Second digit	Level of cloudiness
sunny	1
cirrostratus cloud	2
cloudy	3
very cloudy	4
Third digit	Wind
no wind	0
wind	1
strong wind	2
storm-force wind	3

Third digit	secondary colour
colourless	0
yellowish	1
greenish	2
blueish	3
greyish	4
brownish	5
blackish	6
reddish	7
murky	8

Odour	
odourless	00
chloric odour	01
dirty odour	02
tarry odour	03
oily odour	04
fishy odour	05
decaying, swampy odour	06
odour resembling medicines	07

Ice	
no ice	1
ice along the coastline	2
drift-ice	3
ice cover	4

## 4.11 Conclusion, Choosing the Area

During the preliminary surveys and field inspections, we examined a number of possible sites that could be suitable for a monitoring system. Based on the criteria described above, we can conclude that the 2 km long river section between Letkés and Salka is the most suitable area for our project, and the local competent water management authorities – both Hungarian and Slovak – support our initiative.

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The buoy system to be installed will be able to monitor and detect any possible contamination that gets into the River Ipoly from the Letkés stream as the upper buoy group provides reference values for the lower buoy group – which detects the contamination. Since buoys of the lower buoy group are 200 m away from each other (see Fig. 1), the system can monitor the rate at which the contamination dilutes.



Fig. 1. Location of the buoys and gateway units.

## **5** Implementation

This chapter presents the implementation of the water quality measurement system. First the network topology and the components necessary to achieve the project goals are introduced. Next, the internals of the network elements (buoys) are demonstrated.

#### 5.1 Network Topology

The sensor network created for the WSN-Aqua project is based on data-gathering sensor-equipped buoys placed on the surface of the water. The buoys send their collected data to a Gateway module located on the shore. In case a buoy is not in the radio range of the Gateway, they can send data in multi-hop fashion as they automatically create a mesh-network.

The radio communication between the buoys and the gateway is conducted on 433 MHz ISM band, therefore the nodes have about 500 m line of sight in case of 10 mW output power. This frequency is free to use in the EU for such radio systems, but the communication protocol has to comply with output power and duty cycle restrictions.

The gateway module sends the data collected from the buoys via a cellular connection (EDGE or 3G depending on coverage) to a server module in an aggregated way. However, in case of an emergency (i.e. thresholds are exceeded), the gateway sends the data to the server immediately upon arrival, so the server can send alert messages (SMS and email) to the corresponding authorities.

The management server communicates with the Business Intelligence (BI) server (via internet), which is responsible for gathering statistics and create multiple views and reports of the measured data.

The BI server can send configuration information and new firmware all the way down to the buoys via the private management interface.

The flow of the data in the system is presented in Fig. 2.



Fig. 2. Data flow diagram of the WSN-AQUA project.

#### 5.2 Buoy and Gateway

The buoy (Fig. 3, Fig. 5, Fig. 6) consists of the following components:

- Buoy body (80 x 90 x 50 cm) with lockable opening made from plastic,
- 3 sensors according to Table 4 to gather data,
- Atmel ATxmega128A3 [1] microcontroller to conduct measurements and control the communication,
- TI CC1101 [2] radio module for data transmission,
- 433 MHz antenna for the radio module placed on top of the buoy body for better

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range,

- 5 W solar panel to be able to operate without battery replacement,
- Battery capable of operating for a month without solar power,
- Charge controller board to switch between solar panel and battery power, developed for this project.

Model <sup>8</sup>	Ponsel PHEHT	Ponsel	OPTOD	Ponsel C4E
Measured Property	pH/redox	dissolved	d oxygen	conductivity
Protocol	Modbus RS485 via a	adapter	1	Modbus RS485
Scope	wastewater, surface water, etc.			
Measuring range	0,0014,00 pH, 0 50 °C, -1000+1000 mV	0,00 - 20,0 200% sa	00 mg/l, 0 – aturation	0200 mS/cm, 02000 mS/cm, 020 mS/cm, 0200 mS/cm
Measurement accuracy	$\pm 0,1 \text{ pH}, \pm 0,1 \text{ °C}, \pm 2 \text{ mV}$	± 0,	,1%	$\pm 0,5\%$
Compensation	Temperature	Pressure, te sali	emperature, nity	Temperature
Calibration	5 point, it could be manually calibrated			

Table 4. Sensors used in the WSN-Aqua project.



Fig. 3. Sensor node hardware, embedded ternals.



The Gateway module (Fig. 3) consists of the following components:

- Embedded PC for controlling the module,
- Cellular communication module integrated for sending data to the server,
- TI CC1101 [2] radio module for data transmission with the buoys,
- 433 MHz antenna for the radio module placed on top of the buoy body for better range,
- 100 W solar panel to be able to operate without battery replacement,

<sup>8</sup> www.ponsel-web.com

- Battery capable of operating for a week without solar power,
- Charge controller board to switch between solar panel and battery power, developed for this project.



Fig. 5. The buoy was designed and manufactured in association with Aqua-Terra Lab Ltd., Hungary<sup>9</sup>.



Fig. 6. Deployed buoy during testing.

<sup>9</sup> www.aquaterra.hu

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#### 5.3 **Business Intelligence**

The data gathered from the sensors are sent to a management server via the gateway module. This management server can send alerts in case of an emergency situation (e.g. the thresholds of the measured parameters are exceeded). The data is stored in a database and made available for a data analyzer component. This Business Intelligence server can present the data via a web interface to end users and expert (maintenance) users as well, and is able to collect statistics and generate various reports.

As part of the Business Intelligence, a management system is developed, which is capable of monitoring the status of every node in the network (e.g. battery status, alert configuration, diagnostics, etc.) and initialize firmware upgrade for future requirements. Administrators of the system can monitor and support the system via the management system running on the management server.

#### 6 Research Results

This chapter is an excerpt from [3] and was presented at the Sensornets 2015 conference in Angers, France.

Creating wireless sensor networks requires a different approach than traditional communication networks because energy efficiency plays a key role in sensor networks which consist of devices without external power. The amount of energy used determines the lifetime of these devices. In most cases, data packets are less sensitive to delay, thus can be aggregated, making it possible to gather more useful information reducing the energy required to transmit information.

This chapter discusses the energy efficiency of different Forward Error Correction (FEC) [4] algorithms and presents a method to calculate the optimal amount of aggregation of the data packets in terms of power consumption, while taking into account the Bit Error Rate (BER) characteristics of the wireless channel. In addition, we provide a formula to determine the optimal wakeup signal length, which minimizes the energy consumption.

Fig. 7. shows the gain  $\Theta$  (the ratio of not using aggregation and using it) that can achieved by using aggregation without FEC. The graph line representing BER =  $4 \cdot 10^{-3}$  is jagged because the number of required retransmissions is growing as the



Fig. 7.  $\Theta$  for different BER values without FEC.

aggregation number n is increasing. The number of retransmissions is the same in the neighboring points that follow each other without a jump in their values.

In the figures of this section, the  $\theta = 1$  values are marked with a red line to indicate the level above which the use of aggregation is more efficient.

Remark. This phenomena can be observed in case of other BER values, e.g. for  $BER = 4 \cdot 10^{-5}$  the first jump is at  $n \approx 200$ , which is above the data aggregation value we considered worthy to examine.

Fig. 8. has the same setup as Fig. 7., with the only difference that Hamming [5], [6] codes were applied. The graphs show that in case of medium quality channel  $(BER = 4 \cdot 10^{-4})$  and good quality  $(BER = 4 \cdot 10^{-5})$  channel, there is no difference; the calculated points are perfectly aligned.

Fig. 9 and Fig. 10 show similarity of the values of  $\Theta$  for different BER levels with respect to aggregation number *n* for Reed-Solomon [7], [8] and BCH [9], [10] FEC codes.



Fig. 8.  $\Theta$  for different BER values with Hamming code.

Analyzing Fig. 9 and Fig. 10 it can be noticed, which in case of a poor quality channel ( $BER = 4 \cdot 10^{-3}$ ) for every aggregation number *n* we got better gain  $\theta$  values than in case of better channel. This is because more powerful FEC codes provide more benefits compared to the same aggregation numbers in case of poor quality channels. The better BER channels result in the same gain  $\theta$ .

Also, the graphs look like stages because the block length of Reed-Solomon codes



Fig. 9.  $\Theta$  for different BER values with Reed-Solomon code.

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is fixed. Therefore, if the payload is not long enough, padding is used to fill the rest of the block, which is inefficient.

In Fig. 10 in case of aggregation number n < 40, the poor quality channel gains more using aggregation and BCH code than the better quality channels. Also, at better quality channel, there is significant gain compared to baseline (no aggregation, no FEC) just like in the case of Reed-Solomon codes.



Fig. 10.  $\Theta$  for different BER values with BCH code.

The next three figures (Fig. 11, Fig. 12 and Fig. 13) compare the cases of different FEC codes grouped by channel quality ( $BER = 4 \cdot 10^{-3}, 4 \cdot 10^{-4}$  and  $4 \cdot 10^{-5}$ ) in respect to *n*. For every diagram, a table is included which shows the optimal aggregation number (the highest point of the graphs and the corresponding number of required retransmissions.

Remark. The optimal aggregation number can be much higher is case of BCH and RS codes, but the authors considered n<100 aggregation numbers are worth dealing with because higher aggregation numbers would cause much higher delays. For example, if the aggregation number is n=100 and the packets are generated on an hourly base, then the aggregation delay can be as high as 100 hours. For most real-world scenarios, the delay should be within a day.



Fig. 11. Comparison of FEC codes at BER=4E-3.

Fig. 11 compares FEC codes on the worst quality channel. This scenario shows that the energy cost of different FEC codes is the best. The graph emphasizes that not using any FEC is the worst, and BCH and Reed-Solomon codes perform as the best. It can be seen that in case of lower aggregation numbers (n < 10), Reed-Solomon is the best solution, and from 20 < n < 40 RS and BCH are at the same level. When further increasing the aggregation number, RS code is the most efficient again.

Fig. 12 compares FEC codes on a channel with  $BER = 4 \cdot 10^{-4}$ . It can be seen that in case of n > 20 aggregation numbers, FEC codes provide more energy efficient operation. The FEC codes perform similarly.



Fig. 12. Comparison of FEC codes at BER=4E-4.

According to Fig. 13, in good quality channels there is no benefit of using FEC codes because for every aggregation number, the case without FEC performs the best. The FEC codes just converge to the graph of no FEC case.



Fig. 13. Comparison of FEC codes at BER=4E-5.

#### 7 Conclusions, Future Plans

This book chapter presented the involved partners, implementation details and research results of the WSN-Aqua project. The verification of the proof of concept test system, which is the main output of the project, will be carried out during May to June 2015, therefore the final results cannot be published in this article.

The Middle-Danube Valley Water Directorate is interested in the utilization of the project results with more measured parameters.

Future plans – after the project has been successfully finished – also include further cooperation between the parties involved in new HUSK-CBC tenders.

Acknowledgements. This project has been supported by BME-Infokom Innovator Nonprofit Ltd., http://www.bme-infokom.hu.

This project has been sponsored by The European Union's Hungary-Slovakia Crossborder Co-operation Programme. Building Partnership.

www.husk-cbc.eu, www.hungary-slovakia-cbc.eu

The content of this paper does not necessarily represent the official position of the European Union.



Cross-border Co-operation Programme 2007-2013





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